



EECS 151/251A
Spring 2023
Digital Design and
Integrated Circuits

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Lecture 6:
Combinational Logic
representations

Announcements

- ❑ PS 1 being graded, solution posted tomorrow
- ❑ **PS 2 due Monday**
- ❑ HW 3 out tomorrow.
- ❑ *Wawrzynek office hours moving to Thursday 12:30 starting next week.*

Outline

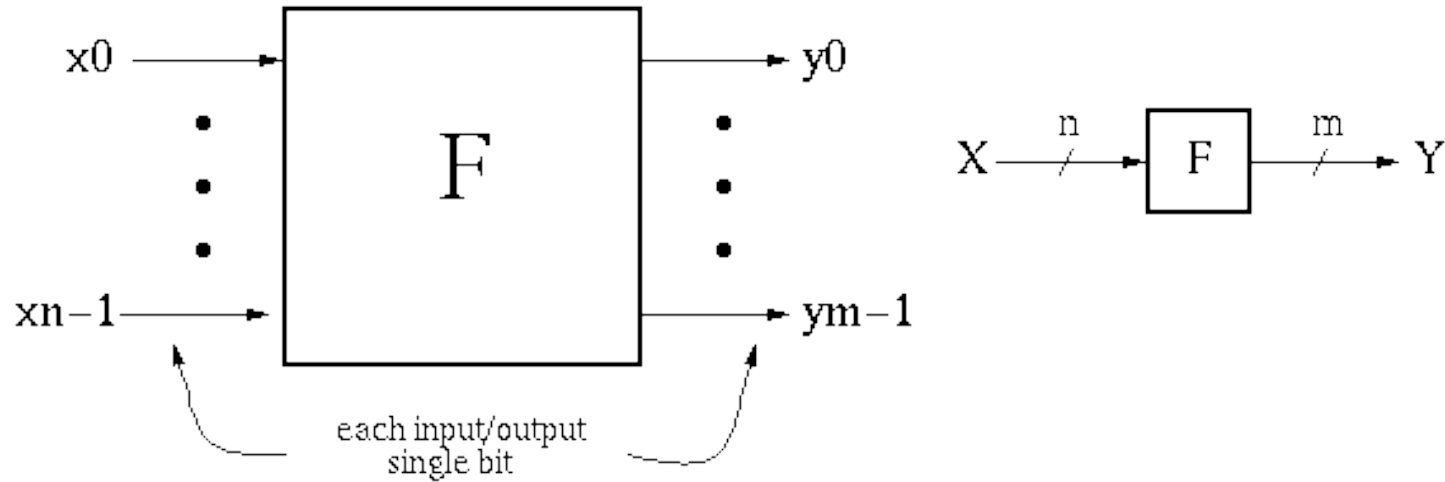


- ❑ *Three representations for combinational logic:*
 - *truth tables,*
 - *graphical (logic gates), and*
 - *algebraic equations*
- ❑ *Boolean Algebra*
- ❑ *Boolean Simplification*
- ❑ *Multi-level Logic,*
- ❑ *NAND/NOR*
- ❑ *XOR*



Representations of Combinational Logic

Combinational Logic (CL) Defined

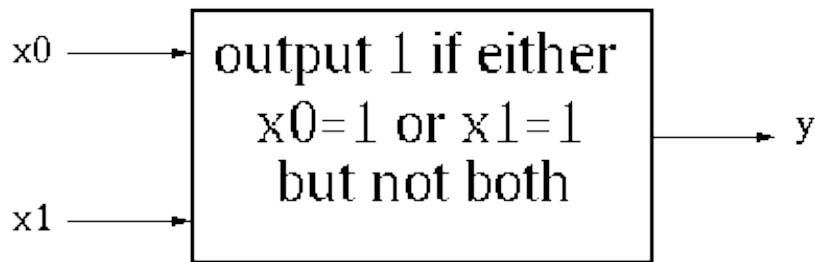


$y_i = f_i(x_0, \dots, x_{n-1})$, where x, y take on values $\{0, 1\}$.

Y is a function of only X , i.e., it is a “pure function”.

- If we change X then Y will change immediately (well almost!).
 - There is an **implementation dependent** delay from X to Y .
 - Y is a function of nothing other than the current inputs values.

CL Block Example #1



Truth Table Description:

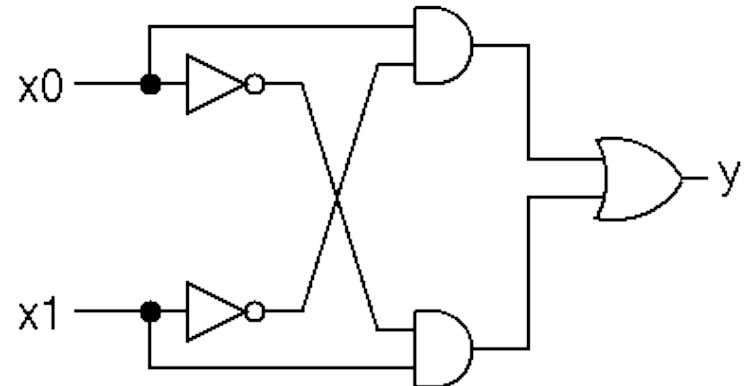
x_0	x_1	y
0	0	0
0	1	1
1	0	1
1	1	0

Boolean Equation:

$$y_0 = (x_0 \text{ AND not}(x_1)) \\ \text{OR } (\text{not}(x_0) \text{ AND } x_1)$$

$$y_0 = x_0x_1' + x_0'x_1$$

Gate Representation:



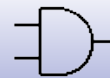
Boolean Algebra/Logic Circuits

❑ Why are they called “logic circuits”?

- ❑ Logic: The study of the principles of reasoning.
- ❑ The 19th Century Mathematician, George Boole, developed a math. system (algebra) involving logic, Boolean Algebra.
- ❑ His variables took on TRUE, FALSE
- ❑ Later Claude Shannon (father of information theory) showed (in his Master's thesis!) how to map Boolean Algebra to digital circuits:
- ❑ Primitive functions of Boolean Algebra:



a	b	AND
0	0	0
0	1	0
1	0	0
1	1	1



a	b	OR
0	0	0
0	1	1
1	0	1
1	1	1



a	NOT
0	1
1	0

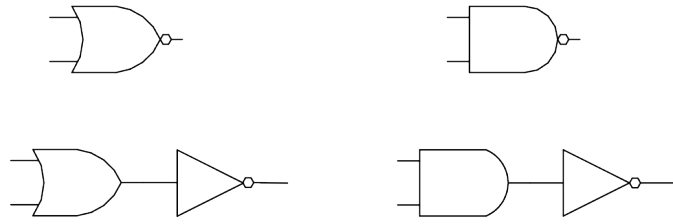


Other logic functions of 2 variables (x,y)

x y	f0	f1														
00	0	0	0	0	0	0	0	0	1	1	1	1	1	1	1	1
01	0	0	0	0	1	1	1	1	0	0	0	0	1	1	1	1
10	0	0	1	1	0	0	1	1	0	0	1	1	0	0	1	1
11	0	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1

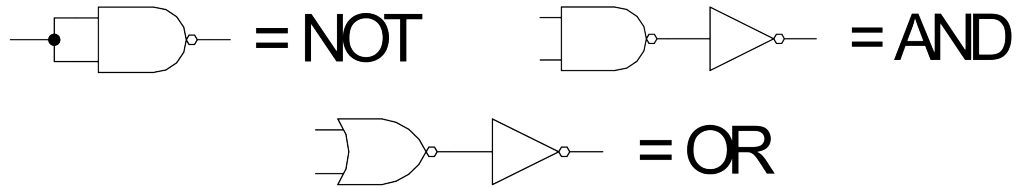
0
AND
X
Y
 \oplus
OR
NOR
XNOR
NAND
1

Look at NOR and NAND:



- Theorem: Any Boolean function that can be expressed as a truth table can be expressed using NAND and NOR.

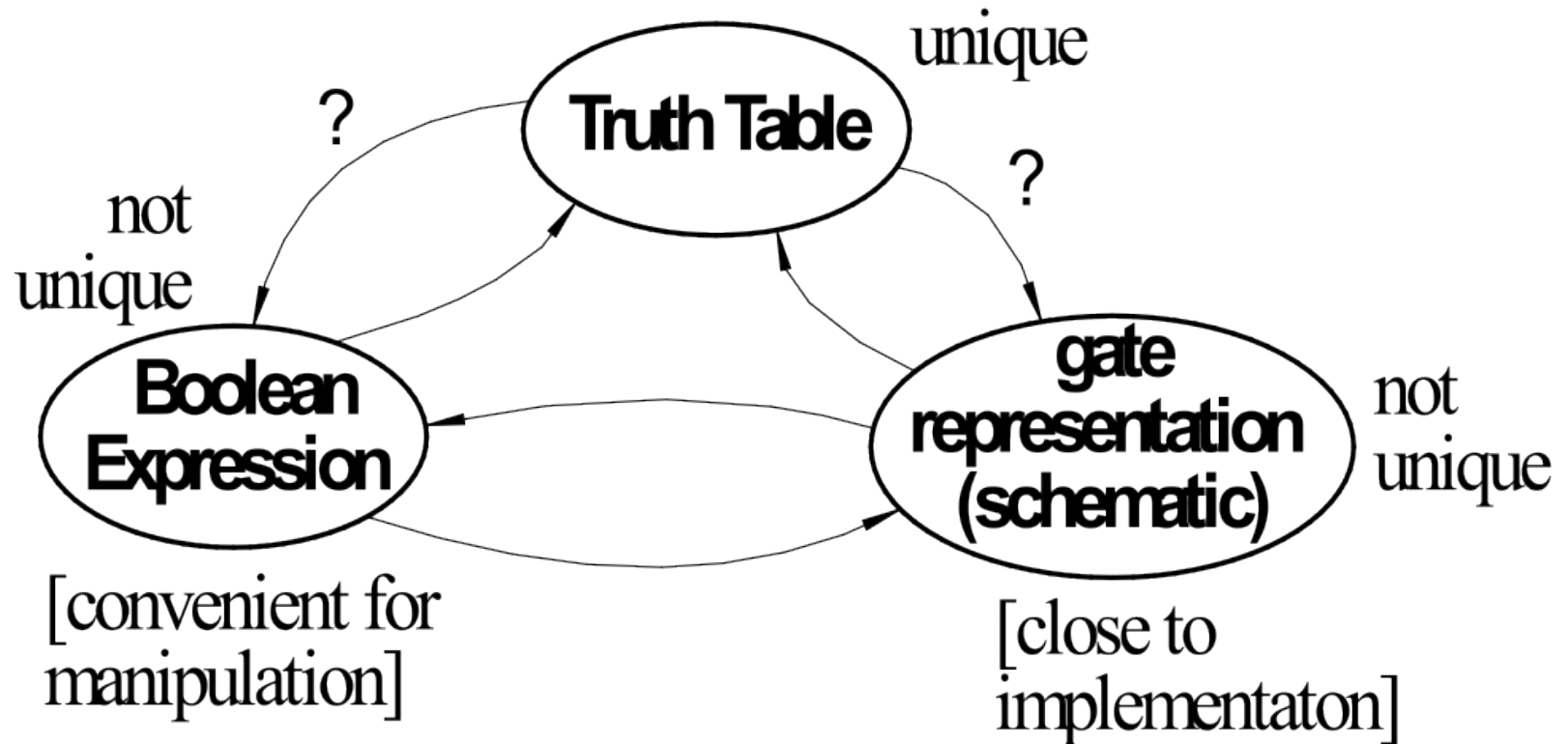
- Proof sketch:



- How would you show that either NAND or NOR is sufficient?

Relationship Among Representations

- * Theorem: Any Boolean function that can be expressed as a truth table can be written as an expression in Boolean Algebra using AND, OR, NOT.



Full-adder (FA) cell example

Graphical Representation of FA-cell

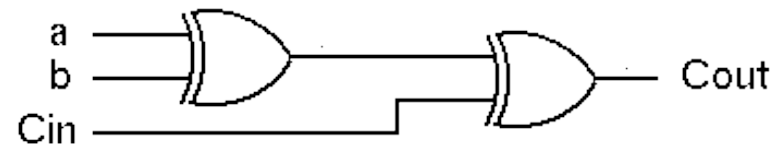
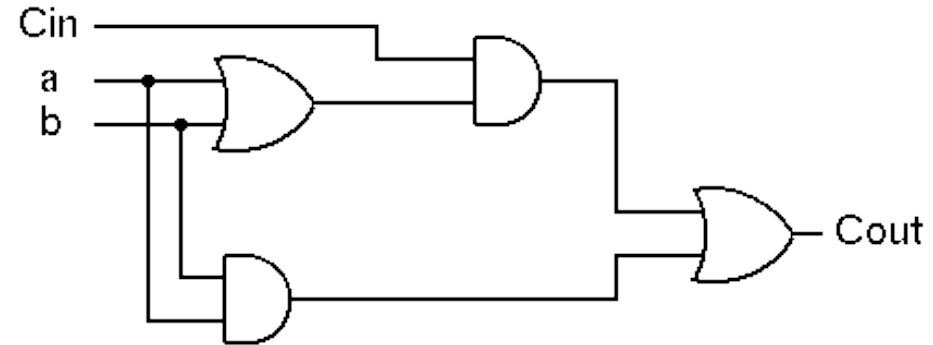
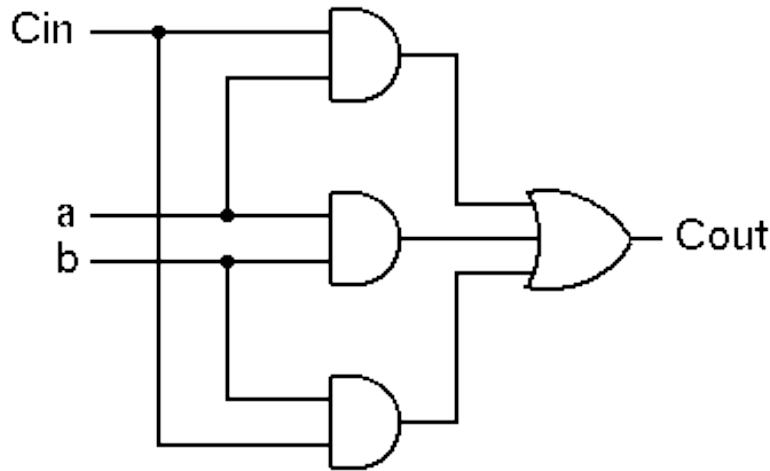
$$r_i = a_i \oplus b_i \oplus c_{in}$$

$$c_{out} = a_i c_{in} + a_i b_i + b_i c_{in}$$

- Alternative Implementation (with only 2-input gates):

$$r_i = [a_i \oplus b_i] \oplus c_{in}$$

$$c_{out} = c_{in}[a_i + b_i] + a_i b_i$$





Boolean Algebra

Boolean Algebra

Set of elements B , binary operators $\{+, \bullet\}$, unary operation $\{'\}$, such that the following axioms hold :

1. B contains at least two elements a, b such that $a \neq b$.

2. Closure : a, b in B ,
 $a + b$ in B , $a \bullet b$ in B , a' in B .

3. Communitive laws :
 $a + b = b + a$, $a \bullet b = b \bullet a$.

4. Identities : $0, 1$ in B
 $a + 0 = a$, $a \bullet 1 = a$.

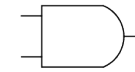
5. Distributive laws :
 $a + (b \bullet c) = (a + b) \bullet (a + c)$, $a \bullet (b + c) = a \bullet b + a \bullet c$.

6. Complement :
 $a + a' = 1$, $a \bullet a' = 0$.

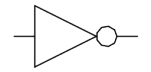
$B = \{0,1\}$, $+$ = OR, \bullet = AND, $'$ = NOT
is a valid Boolean Algebra.



00		0
01		1
10		1
11		1



00		0
01		0
10		0
11		1



0		1
1		0

Some Laws (theorems) of Boolean Algebra

Duality: A dual of a Boolean expression is derived by interchanging OR and AND operations, and interchanging 0s and 1s (the variables are left unchanged).

$$\{F(x_1, x_2, \dots, x_n, 0, 1, +, \bullet)\}^D = \{F(x_1, x_2, \dots, x_n, 1, 0, \bullet, +)\}$$

Any law that is true for an expression is also true for its dual.

Operations with 0 and 1:

$$\begin{array}{ll} \mathbf{x + 0 = x} & \mathbf{x * 1 = x} \\ \mathbf{x + 1 = 1} & \mathbf{x * 0 = 0} \end{array}$$

Idempotent Law:

$$\mathbf{x + x = x} \quad \mathbf{x x = x}$$

Involution Law:

$$\mathbf{(x')' = x}$$

Laws of Complementarity:

$$\mathbf{x + x' = 1} \quad \mathbf{x x' = 0}$$

Commutative Law:

$$\mathbf{x + y = y + x} \quad \mathbf{x y = y x}$$

Some Laws (theorems) of Boolean Algebra (cont.)

Associative Laws:

$$(x + y) + z = x + (y + z)$$

$$x y z = x (y z)$$

Distributive Laws:

$$x (y + z) = (x y) + (x z)$$

$$x +(y z) = (x + y)(x + z)$$

“Simplification” Theorems:

$$x y + x y' = x$$

$$x + x y = x$$

$$x + x'y = x + y$$

$$(x + y) (x + y') = x$$

$$x (x + y) = x$$

$$x(x' + y) = xy$$

DeMorgan's Law:

$$(x + y + z + \dots)' = x'y'z'$$

$$(x y z \dots)' = x' + y' + z'$$

Theorem for Multiplying and Factoring:

$$(x + y) (x' + z) = x z + x' y$$

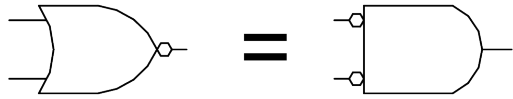
Consensus Theorem:

$$x y + y z + x' z = (x + y) (y + z) (x' + z)$$

$$x y + x' z = (x + y) (x' + z)$$

DeMorgan's Law

$$(x + y)' = x' y'$$



*Exhaustive
Proof*

x	y	x'	y'	(x+y)'	x'y'
0	0	1	1	1	1
0	1	1	0	0	0
1	0	0	1	0	0
1	1	0	0	0	0

$$(x y)' = x' + y'$$

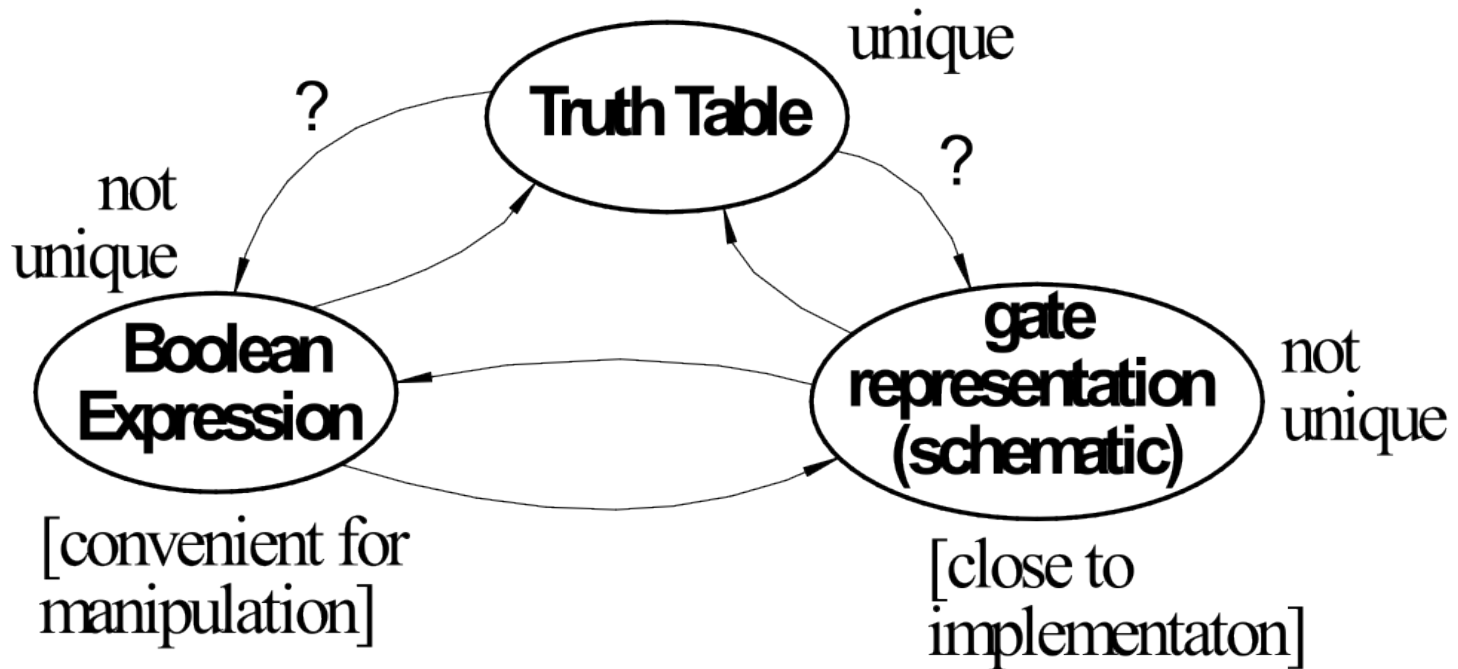


*Exhaustive
Proof*

x	y	x'	y'	(xy)'	x'+y'
0	0	1	1	1	1
0	1	1	0	1	1
1	0	0	1	1	1
1	1	0	0	0	0

Relationship Among Representations

- * Theorem: Any Boolean function that can be expressed as a truth table can be written as an expression in Boolean Algebra using AND, OR, NOT.



How do we convert from one to the other?

Canonical Forms

- Standard form for a Boolean expression - unique algebraic expression directly from a true table (TT) description.
- Two Types:
 - * **Sum of Products (SOP)**
 - * **Product of Sums (POS)**
- Sum of Products (disjunctive normal form, minterm expansion). Example:

Minterms	a	b	c	f	f'
a'b'c'	0	0	0	0	1
a'b'c	0	0	1	0	1
a'bc'	0	1	0	0	1
a'bc	0	1	1	1	0
ab'c'	1	0	0	1	0
ab'c	1	0	1	1	0
abc'	1	1	0	1	0
abc	1	1	1	1	0

One product (and) term for each 1 in f:

$$f = a'bc + ab'c' + ab'c + abc' + abc$$

$$f' = a'b'c' + a'b'c + a'bc'$$

(enumerate all the ways the function could evaluate to 1)

What is the cost?

Sum of Products (cont.)

Canonical Forms are usually not minimal:

Our Example:

$$f = a'bc + ab'c' + ab'c + abc' + abc \quad (xy' + xy = x)$$

$$= a'bc + ab' + ab$$

$$= a'bc + a$$

$$= a + bc$$

$$(x'y + x = y + x)$$

$$f' = a'b'c' + a'b'c + a'bc'$$

$$= a'b' + a'bc'$$

$$= a' (b' + bc')$$

$$= a' (b' + c')$$

$$= a'b' + a'c'$$

Canonical Forms

- Product of Sums (conjunctive normal form, maxterm expansion).

Example:

maxterms	<u>a</u>	<u>b</u>	<u>c</u>	<u>f</u>	<u>f'</u>
$a+b+c$	0	0	0	0	1
$a+b+c'$	0	0	1	0	1
$a+b'+c$	0	1	0	0	1
$a+b'+c'$	0	1	1	1	0
$a'+b+c$	1	0	0	1	0
$a'+b+c'$	1	0	1	1	0
$a'+b'+c$	1	1	0	1	0
$a'+b'+c'$	1	1	1	1	0

One sum (**or**) term for each **0** in f:

$$f = (a+b+c) (a+b+c') (a+b'+c)$$

$$f' = (a+b'+c') (a'+b+c) (a'+b+c') \\ (a'+b'+c) (a+b+c')$$

(enumerate all the ways the function could evaluate to 0)

What is the cost?



Boolean Simplification

Algebraic Simplification Example

Ex: full adder (FA) carry out function (in canonical form):

$$\text{Cout} = a'bc + ab'c + abc' + abc$$

ci	a	b	r	co
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1

Algebraic Simplification

$$\begin{aligned} \text{Cout} &= a'bc + ab'c + abc' + abc \\ &= a'bc + ab'c + abc' + abc + abc \\ &= a'bc + abc + ab'c + abc' + abc \\ &= (a' + a)bc + ab'c + abc' + abc \\ &= (1)bc + ab'c + abc' + abc \\ &= bc + ab'c + abc' + abc + abc \\ &= bc + ab'c + abc + abc' + abc \\ &= bc + a(b' + b)c + abc' + abc \\ &= bc + a(1)c + abc' + abc \\ &= bc + ac + ab(c' + c) \\ &= bc + ac + ab(1) \\ &= bc + ac + ab \end{aligned}$$

Outline for remaining CL Topics

- ❑ K-map method of two-level logic simplification
- ❑ Multi-level Logic
- ❑ NAND/NOR networks
- ❑ EXOR revisited

Algorithmic Two-level Logic Simplification

Key tool: The Uniting Theorem:

$$xy' + xy = x(y' + y) = x(1) = x$$

<i>ab</i>	<i>f</i>
00	0
01	0
10	1
11	1

$$f = ab' + ab = a(b' + b) = a$$

b values change within the on-set rows

a values don't change

b is eliminated, a remains

<i>ab</i>	<i>g</i>
00	1
01	0
10	1
11	0

$$g = a'b' + ab' = (a' + a)b' = b'$$

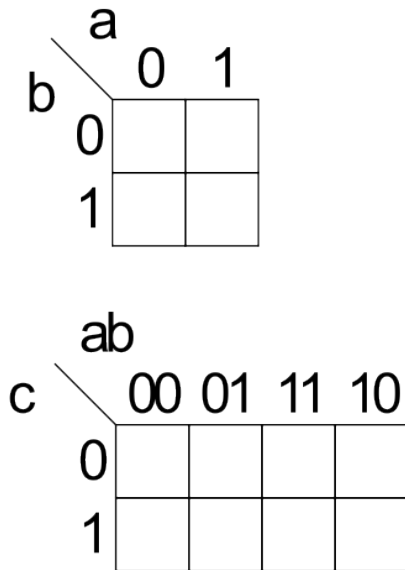
b values stay the same

a values changes

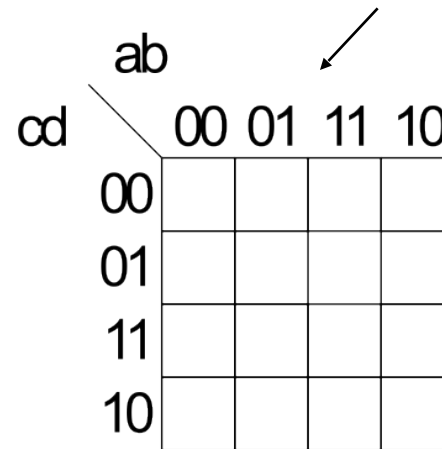
b' remains, a is eliminated

Karnaugh Map Method

- K-map is an alternative method of representing the TT and to help visual the adjacencies.



Note: "gray code" labeling.



5 & 6 variable k-maps possible

Karnaugh Map Method

- Adjacent groups of 1's represent product terms

a

b \ a	0	1
0	0	1
1	0	1

f = a

a

b \ a	0	1
0	1	1
1	0	0

g = b'

ab

c \ ab	00	01	11	10
0	0	0	1	0
1	0	1	1	1

cout = ab + bc + ac

ab

c \ ab	00	01	11	10
0	0	0	1	1
1	0	0	1	1

f = a

K-map Simplification

1. Draw K-map of the appropriate number of variables (between 2 and 6)
2. Fill in map with function values from truth table.
3. Form groups of 1's.
 - ✓ Dimensions of groups must be even powers of two (1x1, 1x2, 1x4, ..., 2x2, 2x4, ...)
 - ✓ Form as large as possible groups and as few groups as possible.
 - ✓ Groups can overlap (this helps make larger groups)
 - ✓ Remember K-map is periodical in all dimensions (groups can cross over edges of map and continue on other side)
4. For each group write a product term.
 - the term includes the “constant” variables (use the uncomplemented variable for a constant 1 and complemented variable for constant 0)
5. Form Boolean expression as sum-of-products.

K-maps (cont.)

		ab			
		00	01	11	10
c	0	1	0	0	1
	1	0	0	1	1

$$f = b'c' + ac$$

		ab			
		00	01	11	10
cd	00	1	0	0	1
	01	0	1	0	0
	11	1	1	1	1
	10	1	1	1	1

$$f = c + a'bd + b'd'$$

(bigger groups are better)

Product-of-Sums K-map

1. Form groups of 0's instead of 1's.
2. For each group write a sum term.
 - the term includes the “constant” variables (use the uncomplemented variable for a constant 0 and complemented variable for constant 1)
3. Form Boolean expression as product-of-sums.

		ab			
		00	01	11	10
cd	00	1	0	0	1
	01	0	1	0	0
	11	1	1	1	1
	10	1	1	1	1

$$f = (b' + c + d)(a' + c + d')(b + c + d')$$

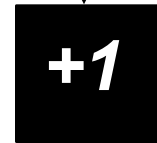
BCD incrementer example

Binary Coded Decimal

	a	b	c	d	w	x	y	z
0	0	0	0	0	0	0	0	1
1	0	0	0	1	0	0	1	0
2	0	0	1	0	0	0	1	1
3	0	0	1	1	0	1	0	0
4	0	1	0	0	0	1	0	1
5	0	1	0	1	0	1	1	0
6	0	1	1	0	0	1	1	1
7	0	1	1	1	1	0	0	0
8	1	0	0	0	1	0	0	1
9	1	0	0	1	0	0	0	0
	1	0	1	0	-	-	-	-
	1	0	1	1	-	-	-	-
	1	1	0	0	-	-	-	-
	1	1	0	1	-	-	-	-
	1	1	1	0	-	-	-	-
	1	1	1	1	-	-	-	-

{a,b,c,d}

4 ↓



4 ↓

{w,x,y,z}

BCD Incrementer Example

- ❑ Note one map for each output variable.
- ❑ Function includes “don’t cares” (shown as “-” in the table).
 - These correspond to places in the function where we don’t care about its value, because we don’t expect some particular input patterns.
 - We are free to assign either 0 or 1 to each don’t care in the function, as a means to increase group sizes.
- ❑ In general, you might choose to write product-of-sums or sum-of-products according to which one leads to a simpler expression.

BCD incrementer example

W

		ab			
cd		00	01	11	10
	00	0	0	-	1
	01	0	0	-	0
	11	0	1	-	-
	10	0	0	-	-

X

		ab			
cd		00	01	11	10
	00	0	1	-	0
	01	0	1	-	0
	11	1	0	-	-
	10	0	1	-	-

w =

x =

y

		ab			
cd		00	01	11	10
	00	0	0	-	0
	01	1	1	-	0
	11	0	0	-	-
	10	1	1	-	-

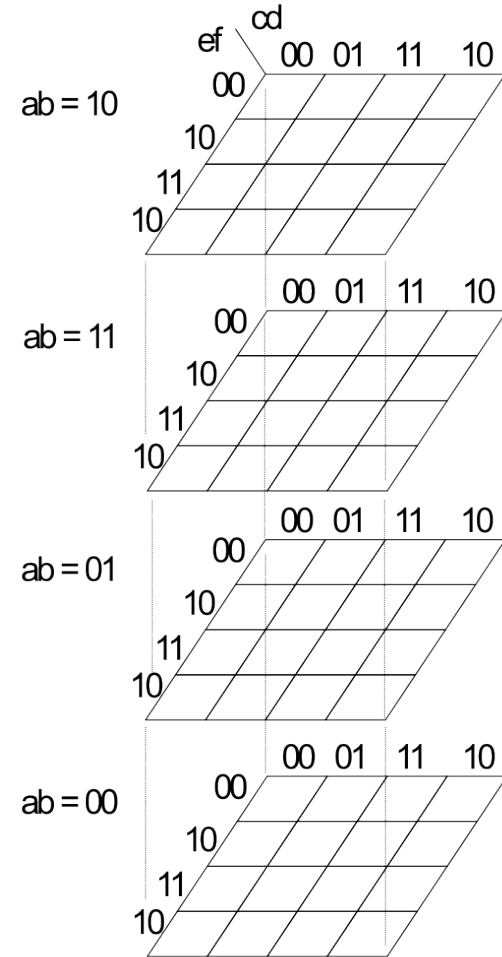
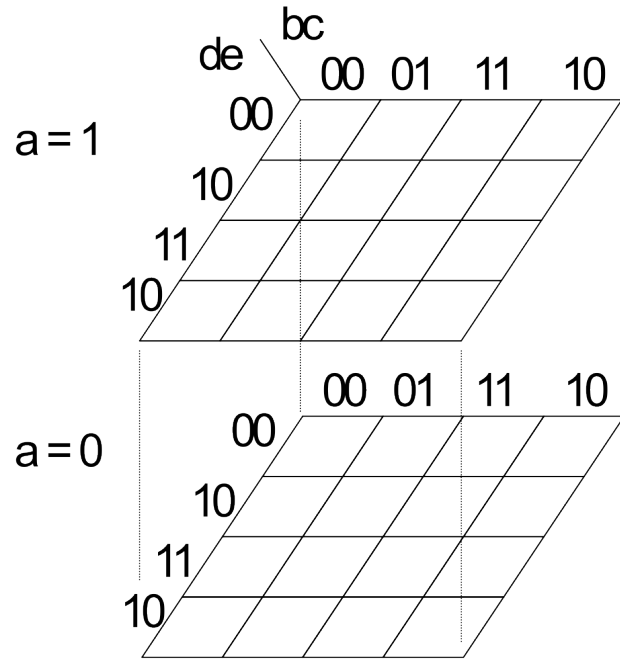
Z

		ab			
cd		00	01	11	10
	00	1	1	-	1
	01	0	0	-	0
	11	0	0	-	-
	10	1	1	-	-

y =

z =

Higher Dimensional K-maps

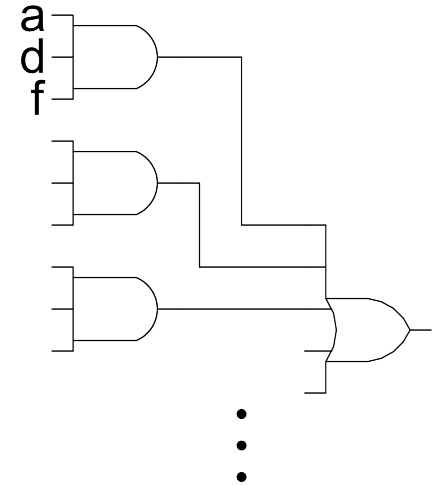




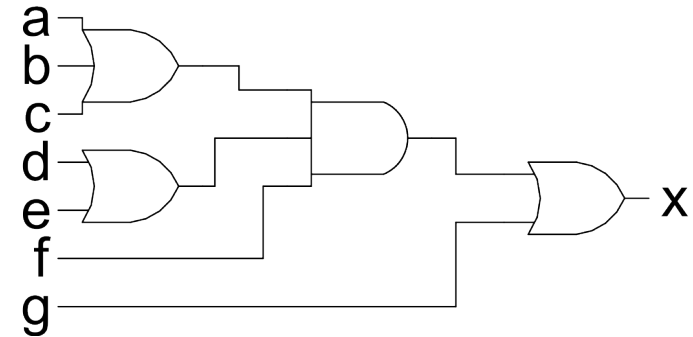
Boolean Simplification – Multi-level Logic

Multi-level Combinational Logic

- Example: reduced sum-of-products form
 $x = adf + aef + bdf + bef + cdf + cef + g$
- Implementation in 2-levels with gates:
cost: 1 7-input OR, 6 3-input AND
=> ~50 transistors
delay: 3-input OR gate delay + 7-input AND gate delay



- Factored form:
 $x = (a + b + c)(d + e)f + g$
cost: 1 3-input OR, 2 2-input OR, 1 3-input AND
=> ~20 transistors
delay: 3-input OR + 3-input AND + 2-input OR



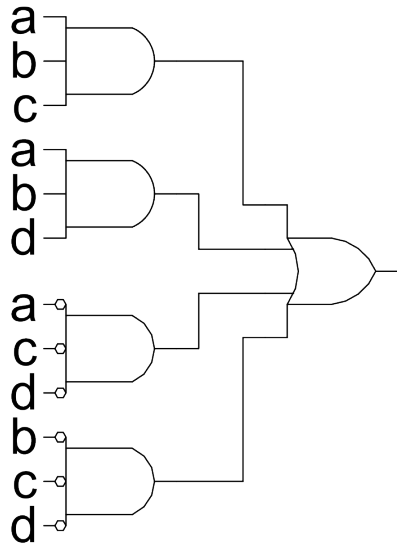
Footnote: NAND would be used in place of all ANDs and ORs.

Which is faster?

*In general: Using multiple levels (more than 2) will reduce the cost. Sometimes also delay.
Sometimes a tradeoff between cost and delay.*

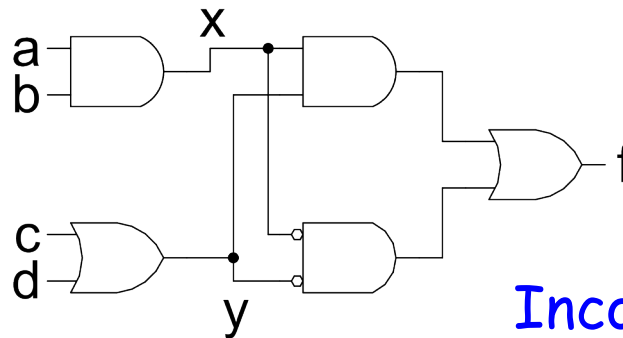
Multi-level Combinational Logic

Another Example: $F = abc + abd + a'c'd' + b'c'd'$



$$\text{let } x = ab \quad y = c+d$$

$$f = xy + x'y'$$



Incorporates fanout.

No convenient hand methods exist for multi-level logic simplification:

a) CAD Tools use sophisticated algorithms and heuristics

Guess what? These problems tend to be NP-complete

b) Humans and tools often exploit some special structure (example adder)

NAND-NAND & NOR-NOR Networks

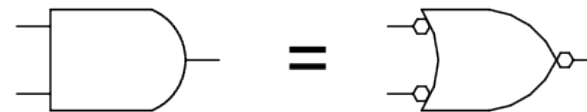
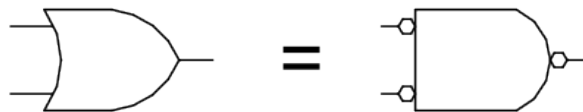
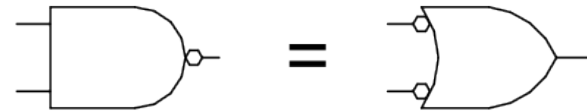
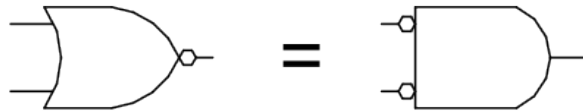
DeMorgan's Law Review:

$$(a + b)' = a' b'$$

$$a + b = (a' b')'$$

$$(a b)' = a' + b'$$

$$(a b) = (a' + b')'$$

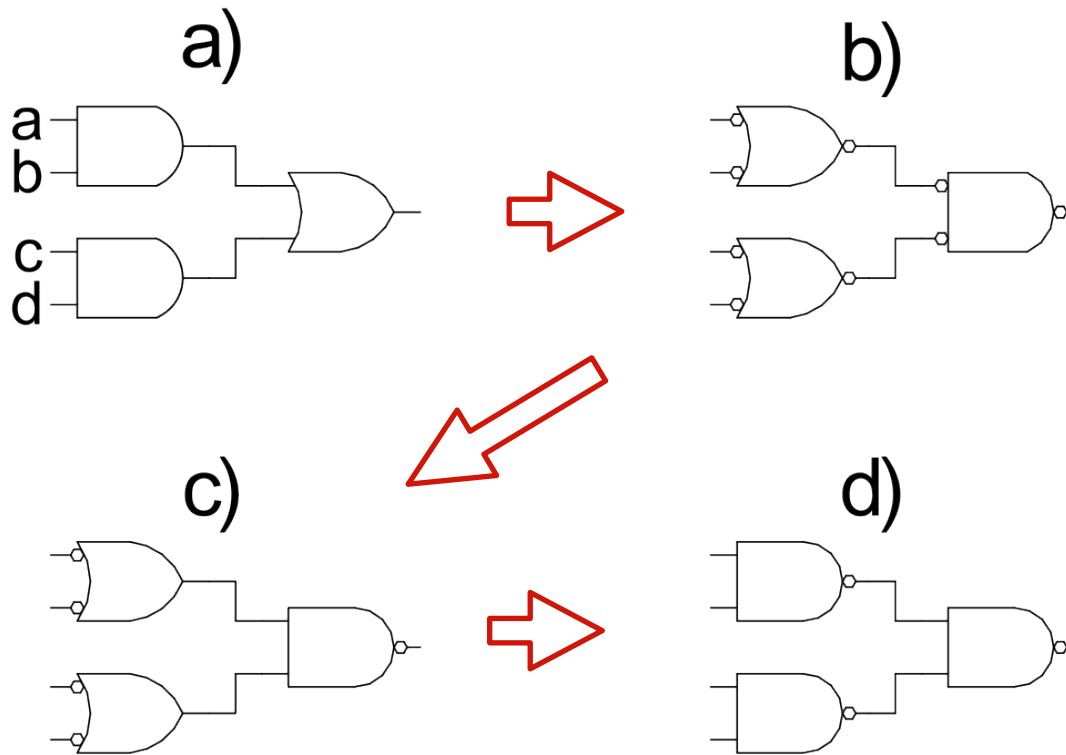


push bubbles or introduce in pairs or remove pairs:

$$(x')' = x.$$

NAND-NAND & NOR-NOR Networks

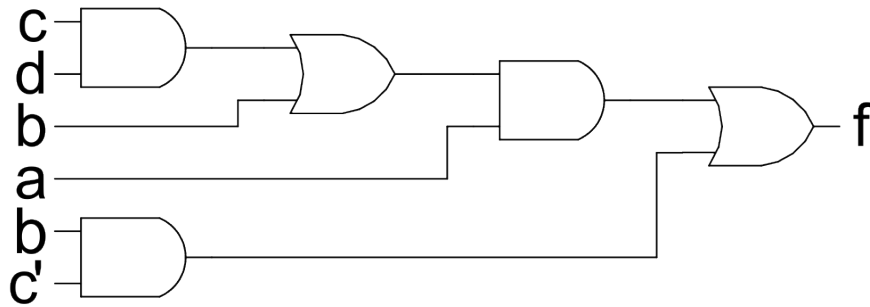
- Mapping from AND/OR to NAND/NAND



Multi-level Networks

Convert to NANDs:

$$F = a(b + cd) + bc'$$

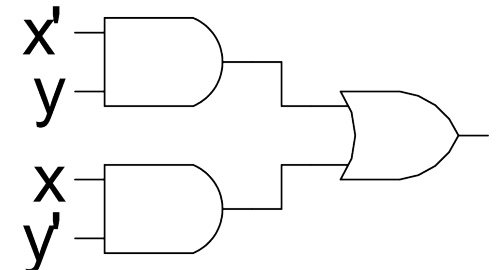
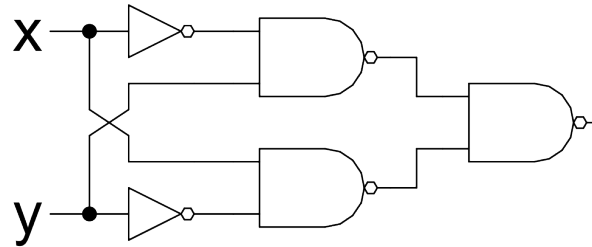


EXOR Function Implementations

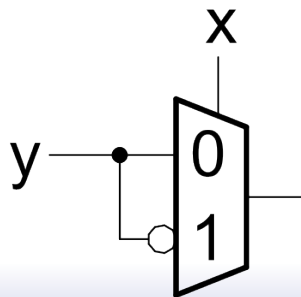
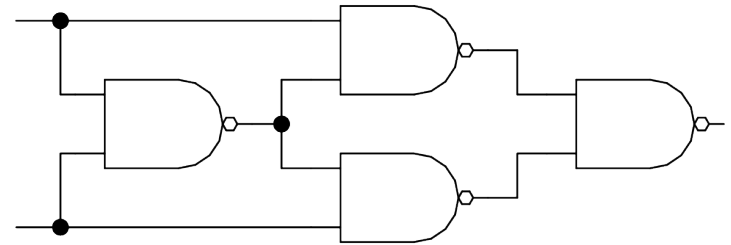
Parity, addition mod 2

$$x \oplus y = x'y + xy'$$

x	y	xor	xnor
0	0	0	1
0	1	1	0
1	0	1	0
1	1	0	1



Another approach:



if $x=0$ then y else y'