# EECS151/251A Discussion 4 

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## Plan for Today

- Show another Verilog trick
- Talk briefly about Boolean Algebra and Optimization
- Answer your questions!
- Experiment!
- Break into groups to do a practice problem
- Reconvene to discuss solution
- Work on a problem as a class


## The casez statement

Making HW2 Problem 2c Easier!

## casez

- You want to use a case block

```
module circuit2_casez(a, y);
    input [3:0] a;
    output reg [1:0] y;
    always @(*)
        casez(a)
        4'b???1: y = 2'b11;
        4'b??10: y = 2'b10;
        4'b?100: y = 2'b01;
        4'b1000: y = 2'b00;
        default: y = a[1:0];
        endcase
endmodule
```

Boolean Algebra

## Boolean Algebra: A mathematical way of looking at logic

- Basic operators: AND ( $\cdot, \wedge$ ), OR (+, V), NOT ( $\neg$, , , !, ~, or "bar" - ex: $\bar{a}$ )
- Like standard algebra, AND (•) takes precedence over OR (+). NOT takes precedence over AND
- $a^{\prime} b+b c=\left(\left(a^{\prime}\right) \cdot b\right)+(b \cdot c)$
- Like standard algebra, there are a set of laws that can be applied to Boolean expressions
- We can use these laws to simplify expressions


## Important Properties

- Many properties are listed in Lecture 6 Slides
- Make a note of these properties, they will be useful!
- Here is a short summary of some (but not all) of them:

| $a b=b c$ | $a+b=b+a$ | $a^{\prime \prime}=a$ |  |
| :---: | :---: | :---: | :---: |
| $(a b) c=a(b c)$ | $(a+b)+c=a+(b+c)$ | $a \cdot 0=0$ | $a+1=1$ |
| $a(b+c)=a b+a c$ | $a+b c=(a+b)(a+c)$ | $a \cdot 1=a$ | $a+0=a$ |
| $(a+b+\cdots+c)^{\prime}=a^{\prime} b^{\prime} \cdots c^{\prime}$ | $(a b \cdots c)^{\prime}=a^{\prime}+b^{\prime}+\cdots+c^{\prime}$ | $a \cdot a=a$ | $a+a=a$ |
| $a b^{\prime}+a b=a\left(b^{\prime}+b\right)=a(1)=a$ | $a \cdot \bar{a}=0$ | $a+\bar{a}=1$ |  |

## Useful Tricks

- You can introduce redundant terms
- $A B+B C=A B+B C+B C$
- Why would you want to do this? To introduce terms from which you can factor


## Canonical Forms

- A Boolean expression can be converted to a set of canonical forms
- Sum of Product (SOP) is one canonical form consisting of the sum (OR) of a series of products (ANDs)
- Ex: $a b c+c d+e f g h$
- Product of Sum (POS) is another canonical form consisting of the product (AND) of a series of sums (ORs)
- Ex: $(x+y+z)(z+h)(h+i+j+k)$
- There are methods to derive the optimal 2 level logic expression (in SOP or POS form)
- K-maps are a tool which we can use by hand to find these simplified expressions


## K-Maps

- Method by which we can more easily observe adjacencies in the truth table
- By constructing large rectangles that are even powers of 2 , we can derive the minimal SOP or POS expression

| $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 00 |  |  |  |  |
| 01 |  |  |  |  |
| 11 |  |  |  |  |
| 10 |  |  |  |  |

Example K-Map for 4 Input Expression

## K-Maps

|  | $A^{\prime} B^{\prime}$ | $A^{\prime} B$ | $A B$ | $A B^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| 00 |  |  |  |  |
| 01 |  |  |  |  |
| 11 |  |  |  |  |
| 10 |  |  |  |  |


|  | $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $C^{\prime} D^{\prime}$ | 00 |  |  |  |  |
| $C^{\prime} D$ | 01 |  |  |  |  |
| $C D$ | 11 |  |  |  |  |
| $C D^{\prime}$ | 10 |  |  |  |  |

## K-Maps

|  | $A^{\prime}$ |  | $A$ |  |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| 00 |  |  |  |  |
| 01 |  |  |  |  |
| 11 |  |  |  |  |
| 10 |  |  |  |  |


|  | $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $C^{\prime}$ | 00 |  |  |  |  |
|  | 01 |  |  |  |  |
| $C$ | 11 |  |  |  |  |
|  | 10 |  |  |  |  |


| $B^{\prime}$ |  | $B$ |  | $B^{\prime}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| 00 |  |  |  |  |
| 01 |  |  |  |  |
| 11 |  |  |  |  |
| 10 |  |  |  |  |


|  | $\mathrm{CD} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $D^{\prime}$ | 00 |  |  |  |  |
| $D$ | 01 |  |  |  |  |
|  | 11 |  |  |  |  |
| $\mathrm{D}^{\prime}$ | 10 |  |  |  |  |

Finite State Machines (FSMs)

## Finite State Machines (FSMs)

- Allows us to design/model complex systems by viewing a system as having a set of possible states it can be in
- The machine can only be in one state at a time
- There are rules dictating how the machine moves between states
- The output is either based solely on the current state (Moore style), the current state and current inputs (Mealy style), or a combination of these
- Very common in digital logic
- Often used to design "control logic"
- ASIC and FPGA labs will both be using FSMs like this
- So common that many EDA tools (including Vivado) have special optimization passes specifically for FSMs
- Can also be used in software, particularly in Real-Time Systems \& Mechatronics


## Group Work

## Problem 1

- Simplify the following expressions using a k-map:
module circuit1(a, b, c, y z)
input a, b, c;
output $y, z$;

$$
\begin{aligned}
& \quad \begin{array}{l}
\text { assign } y=a \& b \& c|a \& b \& \sim c| a \& \sim b \& c ; \\
\text { assign } z=a \& b \mid \sim a \& \sim b ; \\
\text { endmodule }
\end{array}
\end{aligned}
$$

- Use Boolean Algebra to transform the original expression into the simplified one from your k-map


## Problem 1

## y

|  |  | $A^{\prime}$ |  | $A$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $A^{\prime} B^{\prime}$ | $A^{\prime} B$ | $A B$ | $A B^{\prime}$ |
|  | $\mathrm{C} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| $C^{\prime}$ | 0 |  |  | 1 |  |
| $C$ | 1 |  |  | 1 | 1 |

Simplified Expr: $A B+A C$

$$
A B C+A B C^{\prime}+A B^{\prime} C
$$

$$
=A B C+A B C^{\prime}+A B C+A B^{\prime} C
$$

$$
=A B\left(C+C^{\prime}\right)+A C\left(B+B^{\prime}\right)
$$

$=A B(1)+A C(1)$
$=A B+A C$

|  |  | $A^{\prime}$ |  | $A$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $A^{\prime} B^{\prime}$ | $A^{\prime} B$ | $A B$ | $A B^{\prime}$ |
|  | $\mathrm{C} \backslash \mathrm{AB}$ | 00 | 01 | 11 | 10 |
| $C^{\prime}$ | 0 | 1 |  | 1 |  |
| $C$ | 1 | 1 |  | 1 |  |

Simplified Expr: $A B+A^{\prime} B^{\prime}$ Unchanged

