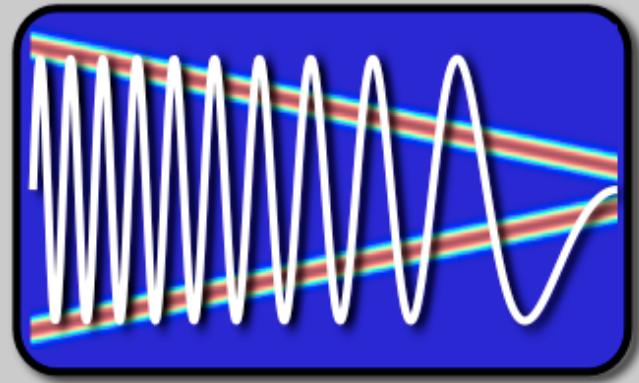


EE123

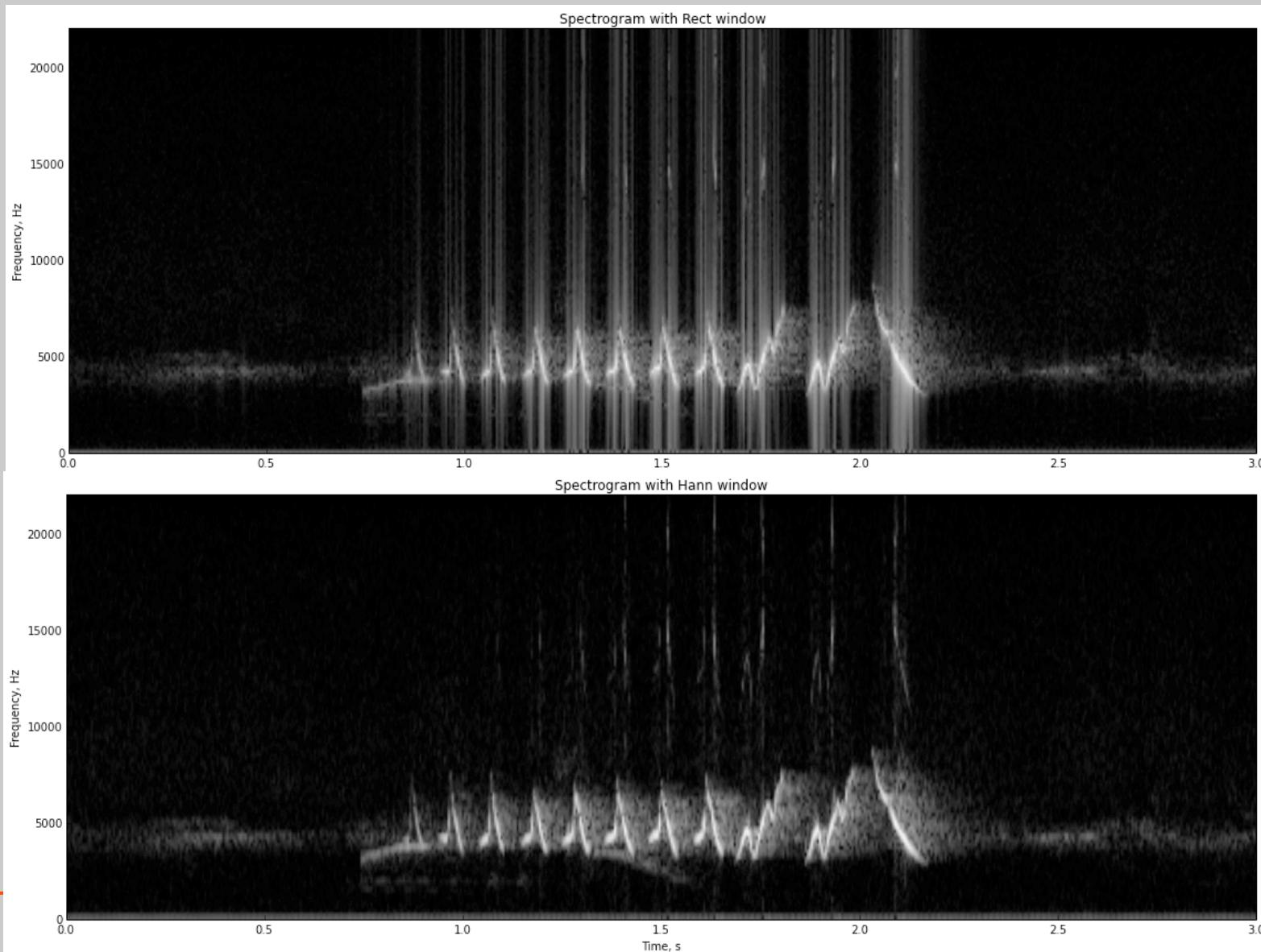


Digital Signal Processing

Lecture 16

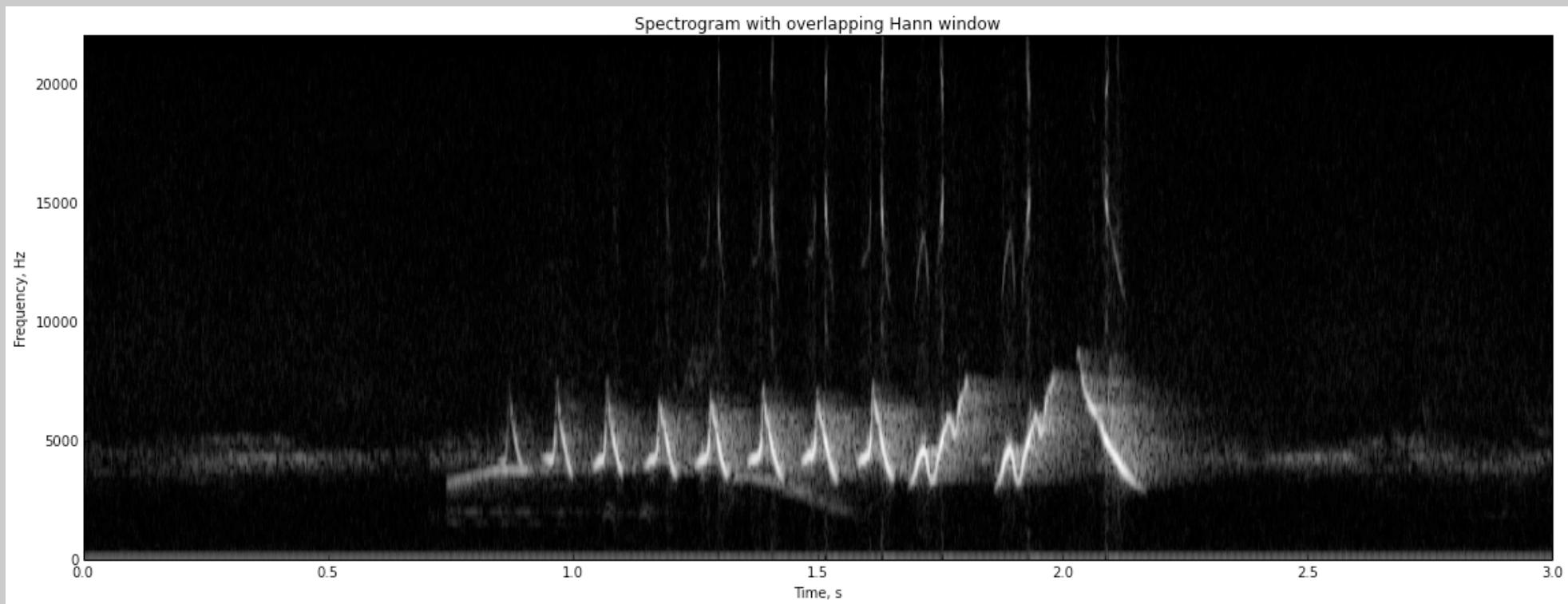
Lab II - Time-Frequency

- compute spectrograms with w/o windowing



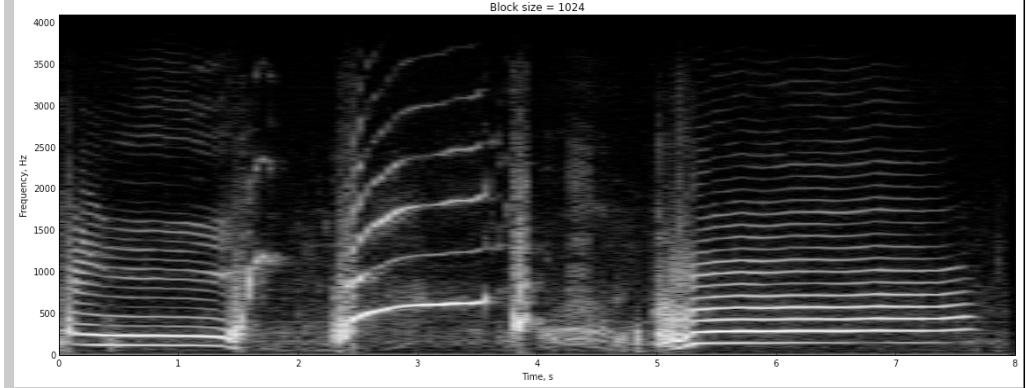
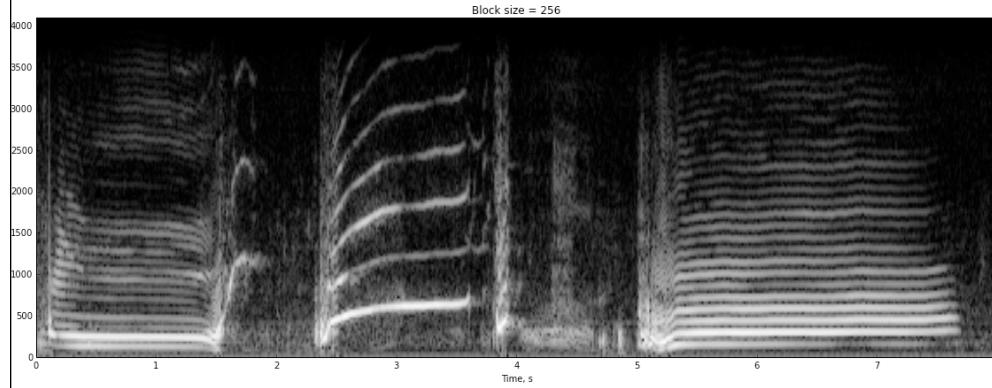
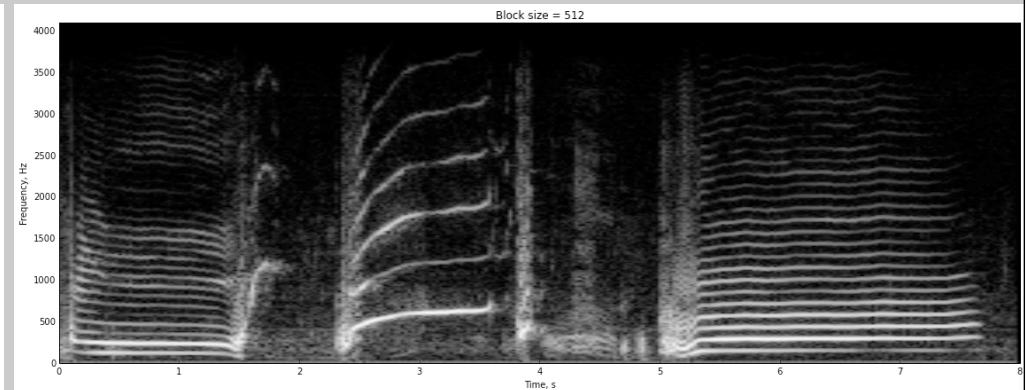
Lab II - Time-Frequency

- Compute with overlapping window

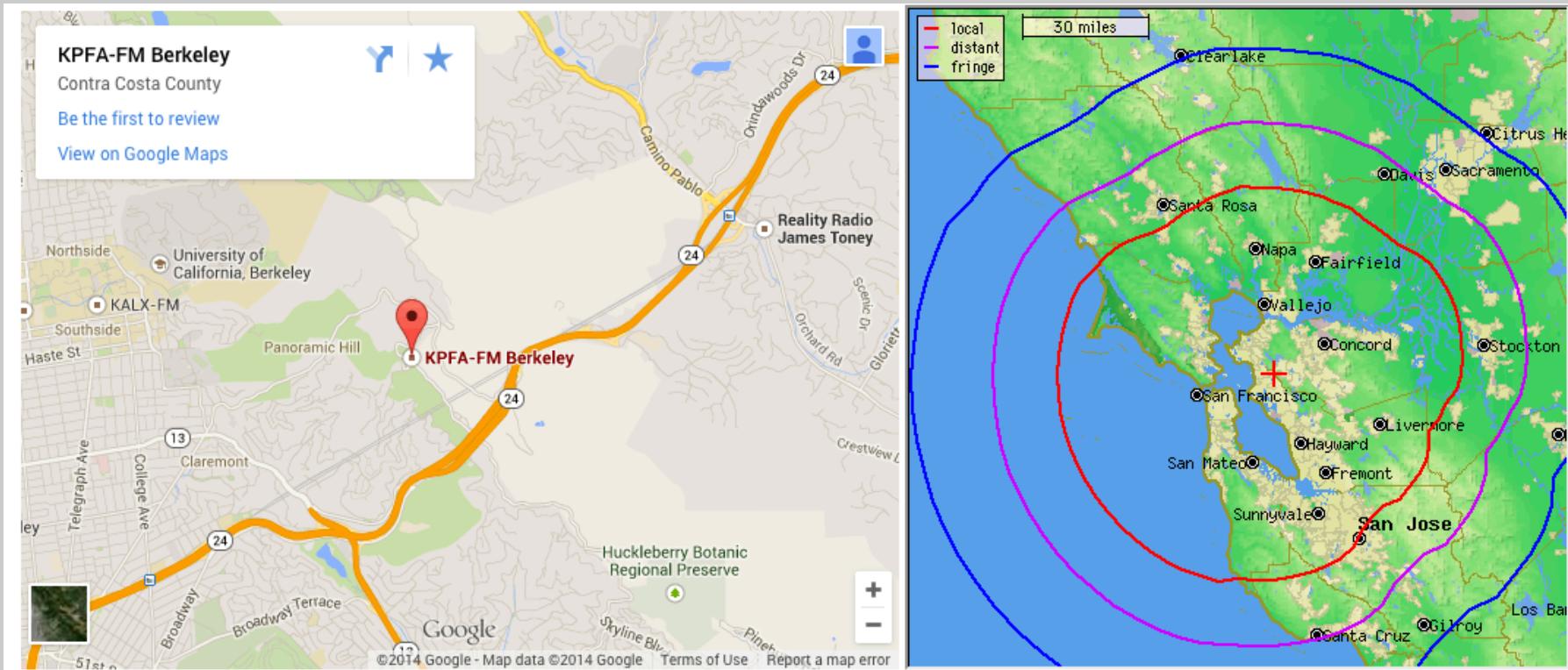


Lab II - Time-Frequency

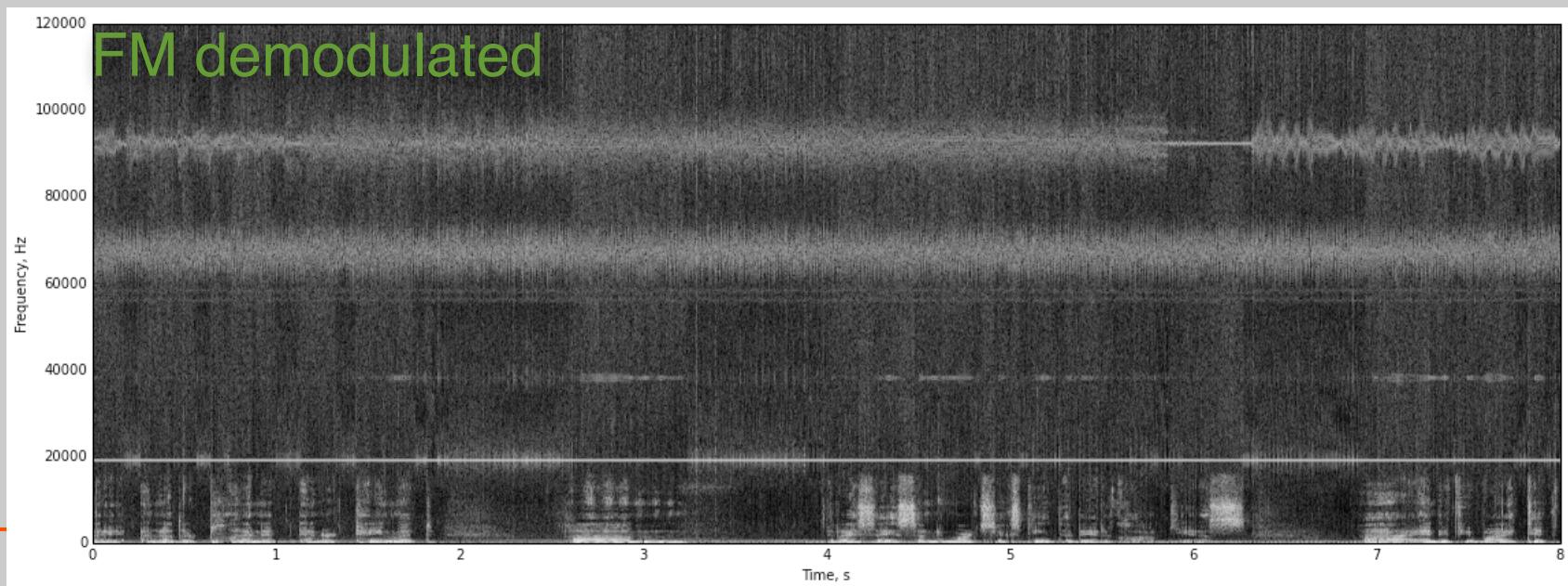
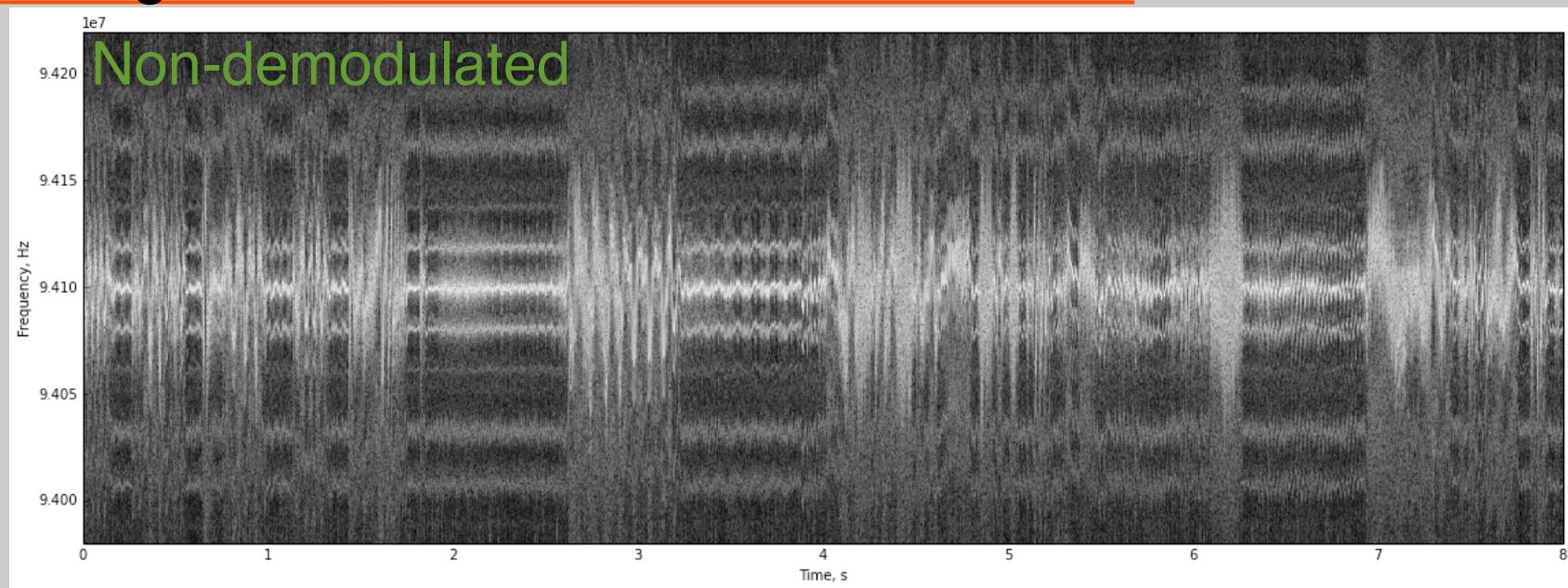
- Look at temporal/frequency resolution tradeoffs:



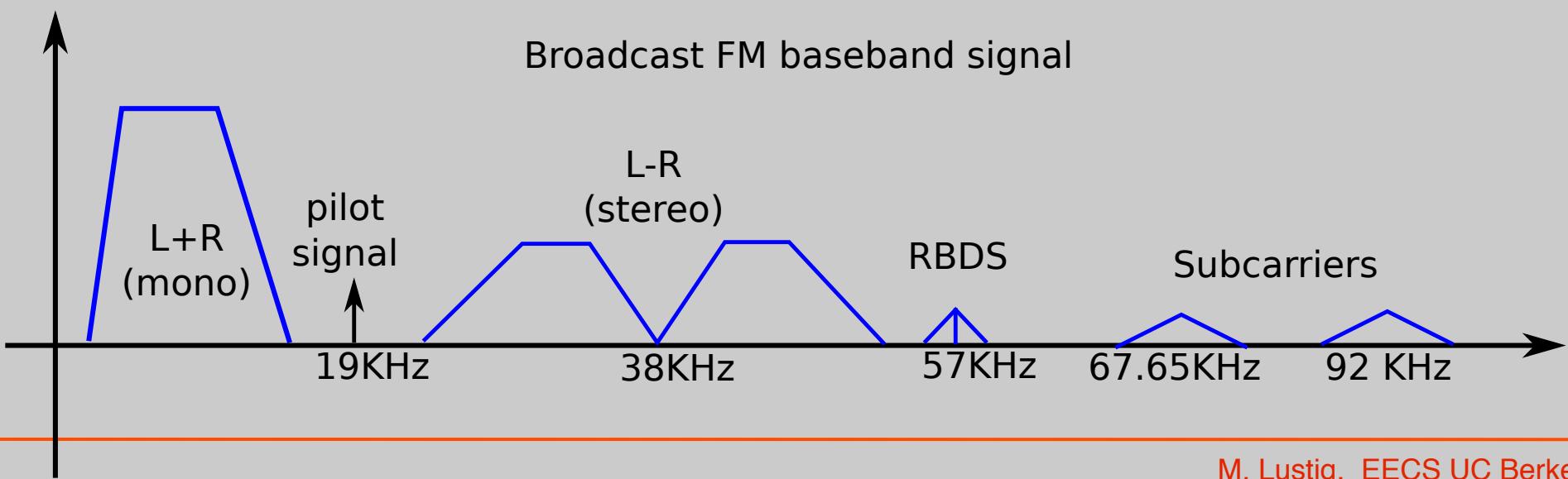
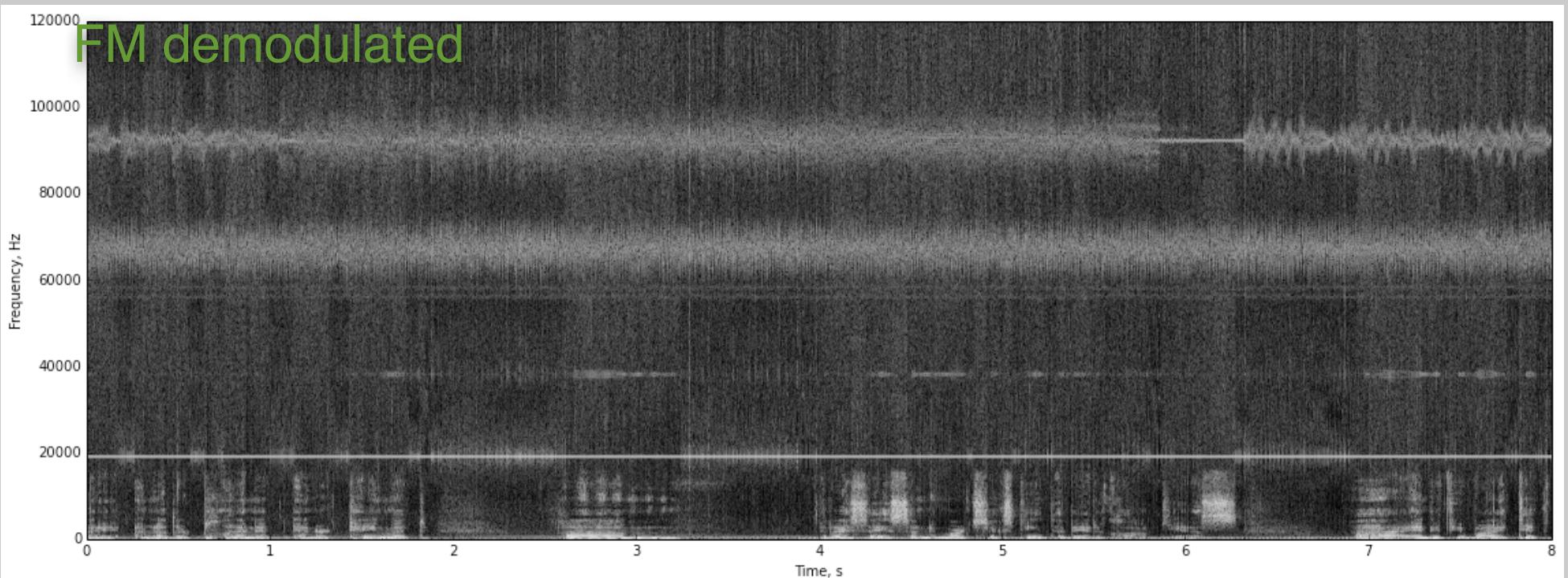
FM Broadcast Radio - KPFA 94.1MHz



Spectrogram of Broadcast FM



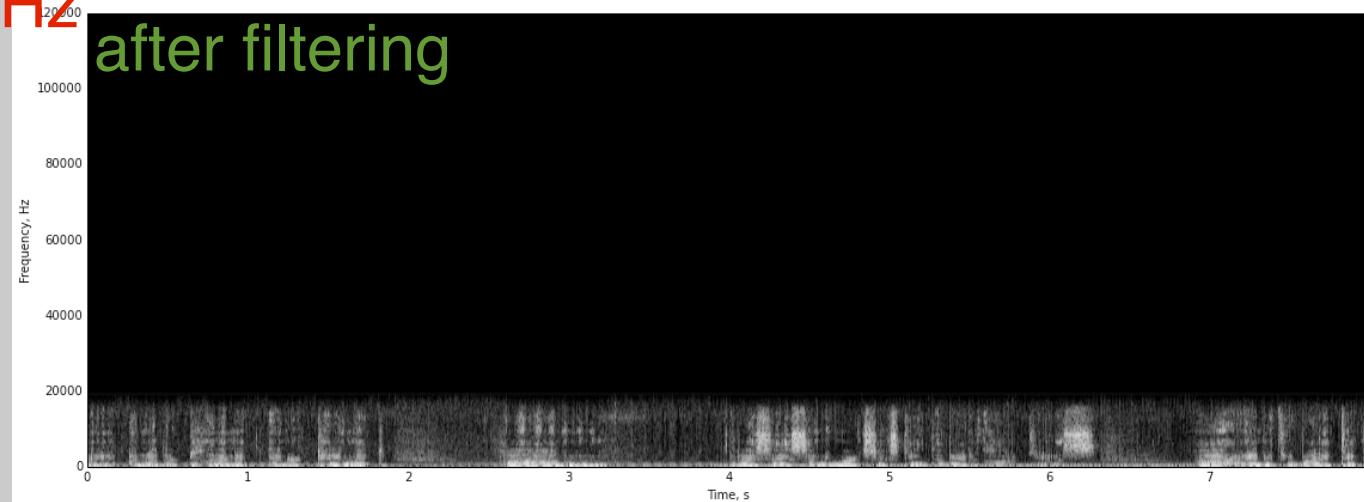
Spectrogram of Broadcast FM



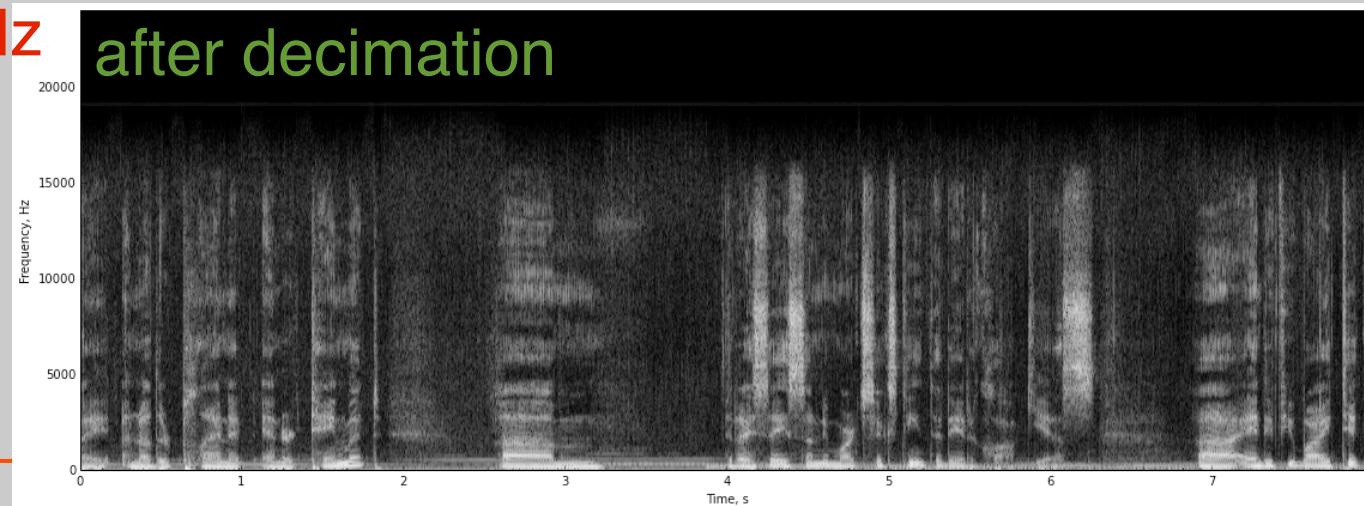
Filter Mono and down

- To play we need to filter the right signal
- Downsample to 48KHz so we can play on the computer

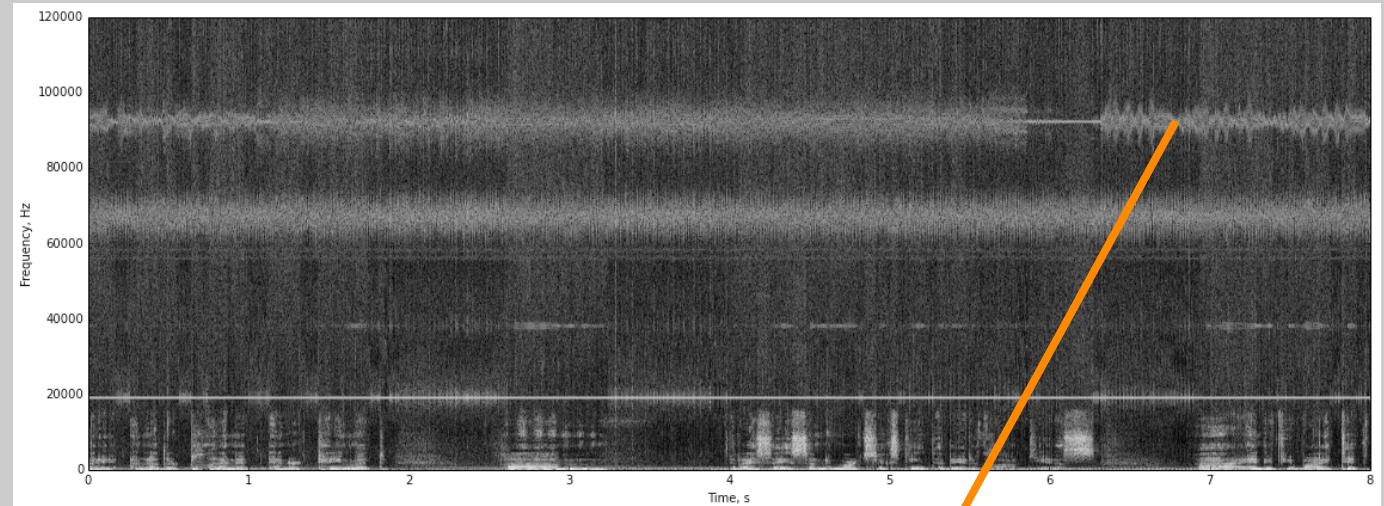
120KHz



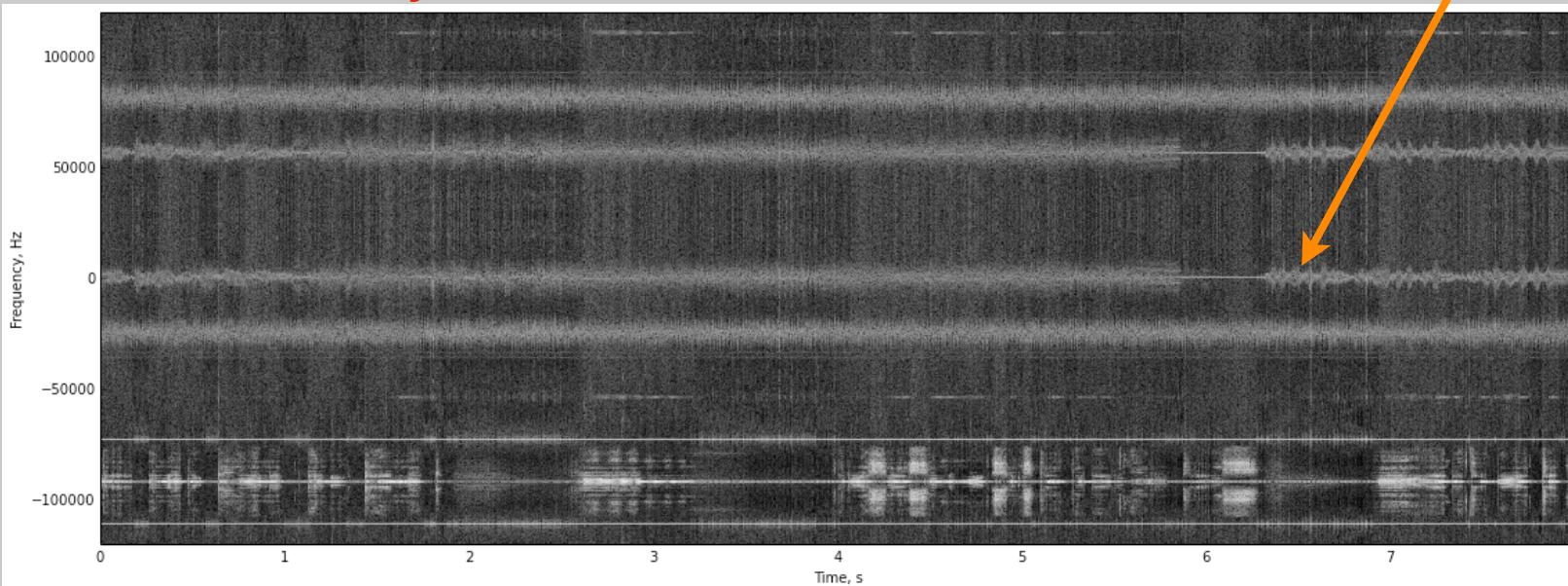
24KHz



Demodulate subcarriers: Example 92KHz

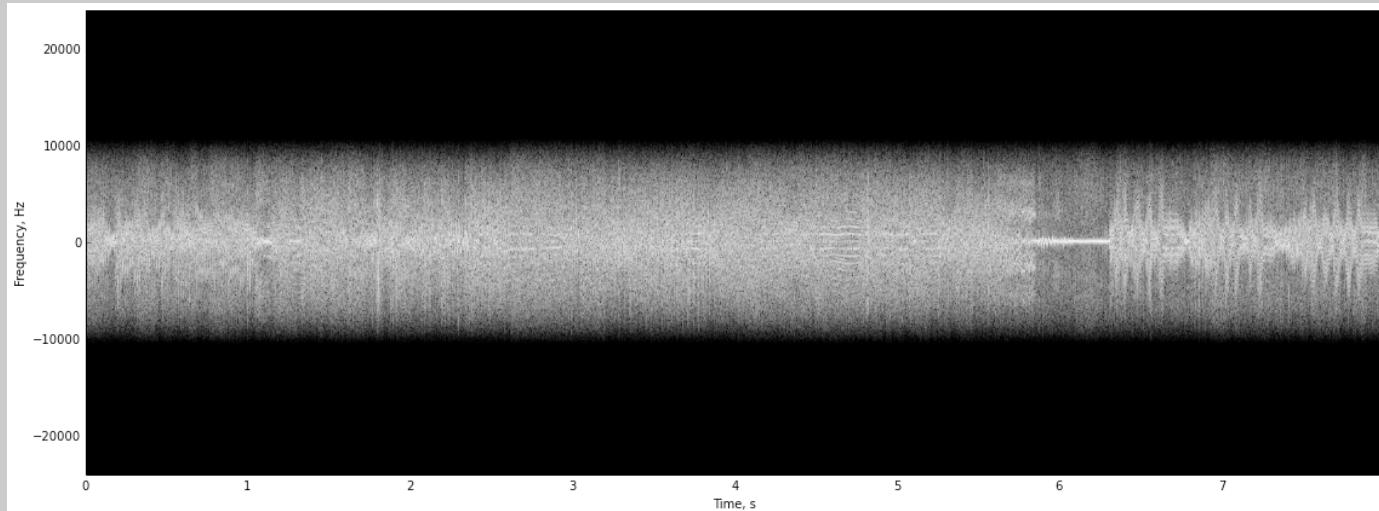


demodulate by 92KHz

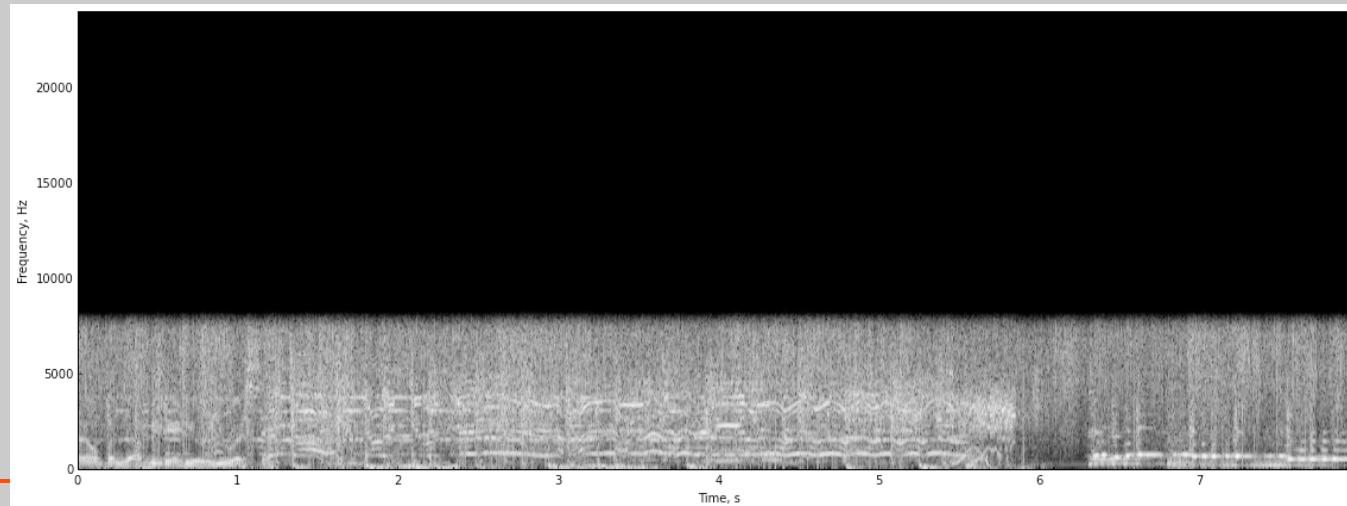


Demodulate subcarriers: Example 92KHz

- Filter and decimate



- FM demodulate and filter



Topics

- Last time
 - Ideal reconstruction D/C
 - D.T processing of C.T signals
 - C.T processing of D.T signals (ha?????)
- Today
 - Changing Sampling Rate via DSP
 - Downsampling
 - Upsampling

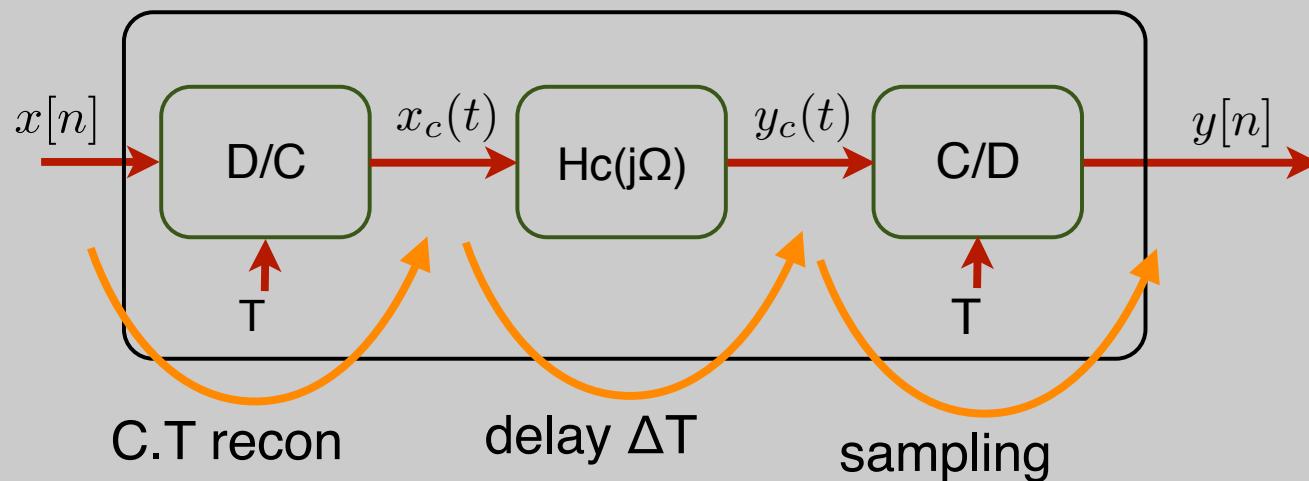
Example:

Non-integer delay:

$$H(e^{j\omega}) = e^{-j\omega\Delta}$$

- What is the time-domain operation when Δ is not an integer ($\Delta=1/2$)?

Let: $H_c(j\Omega) = e^{-j\Omega\Delta T}$ delay of ΔT in time



Example: Non Integer Delay

- The block diagram is only for interpretation!

$$y_c(t) = x_c(t - \Delta)$$

$$\begin{aligned} y[n] &= y_c(nT) = x_c(nT - T\Delta) \\ &= \sum_k x[k] \text{sinc} \left(\frac{t - kT - T\Delta}{T} \right) \Big|_{t=nT} \end{aligned}$$

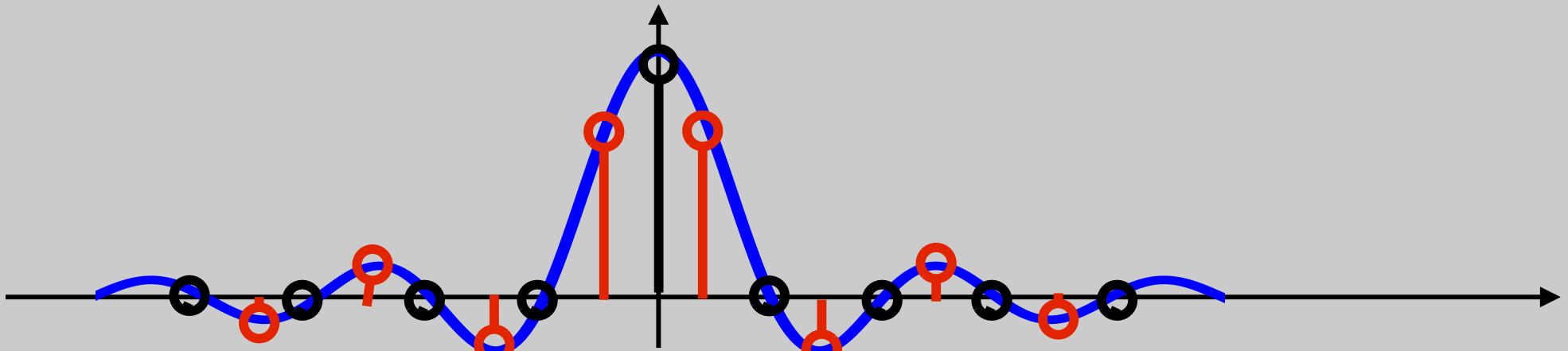
T's cancel!

$$= \sum_k x[k] \text{sinc}(n - k - \Delta)$$

Example: Non Integer Delay

$$h[n] = \text{sinc}(n - \Delta)$$

Example: a discrete delta is a representation of a sampled sinc



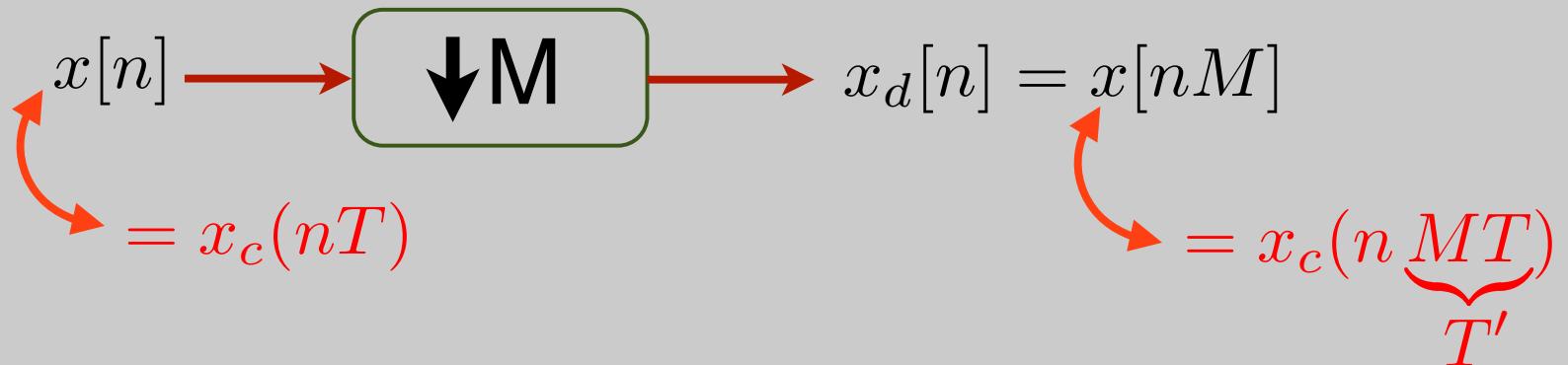
shifted by partial samples results in many coefficients!

DownSampling

- Much like C/D conversion
- Expect similar effects:
 - Aliasing
 - mitigate by antialiasing filter
- Finely sampled signal \Rightarrow almost continuous
 - Downsample in that case is like sampling!

Changing Sampling-rate via D.T Processing

Downsampling:



The DTFT:

$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}}_{\Omega} - \underbrace{\frac{2\pi}{T} k}_{\Omega_s} \right) \right)$$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

Changing Sampling-rate via D.T Processing

The DTFT:

$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}}_{\Omega} - \underbrace{\frac{2\pi}{T} k}_{\Omega_s} \right) \right)$$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

we would like to bypass X_c and go from $X(e^{j\omega}) \Rightarrow X_d(e^{j\omega})$

substitute $r = kM + i$ $i=0, 1, \dots, M-1$
 $k=-\infty, \dots, \infty$

two counters

e.g., k : hours, i : minutes

Changing Sampling-rate via D.T Processing

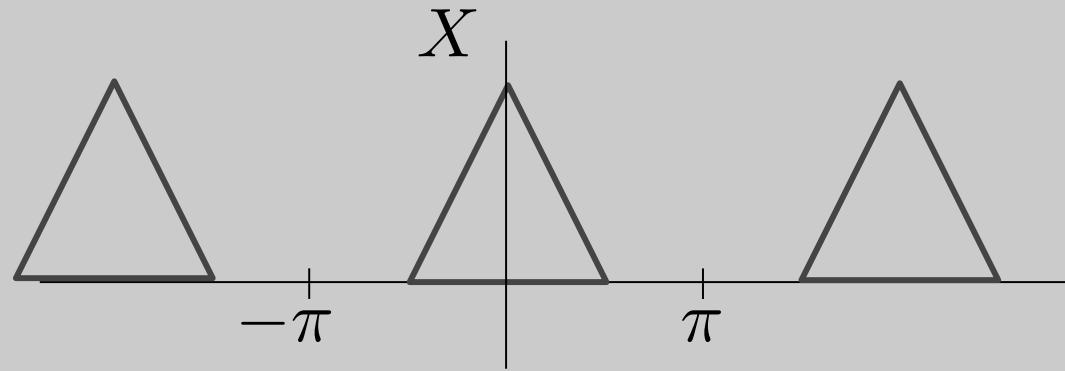
$$\begin{aligned} X_d(e^{j\omega}) &= \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right) \\ &= \frac{1}{M} \sum_{i=0}^{M-1} \underbrace{\frac{1}{T} \sum_{-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} i - \frac{2\pi}{T} k \right) \right)}_{X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})} \end{aligned}$$

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})$$

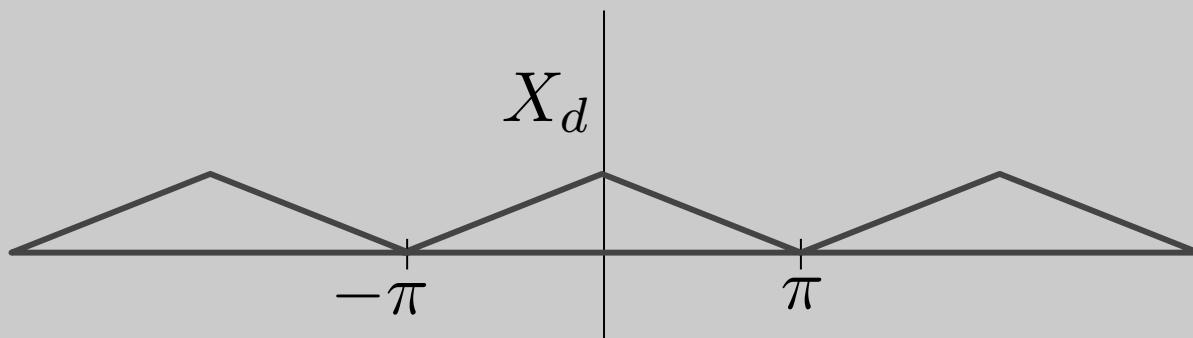
↑ stretch by M ↑ replicate

Changing Sampling-rate via D.T Processing

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0} X \left(e^{j(\omega/M - 2\pi i/M)} \right)$$

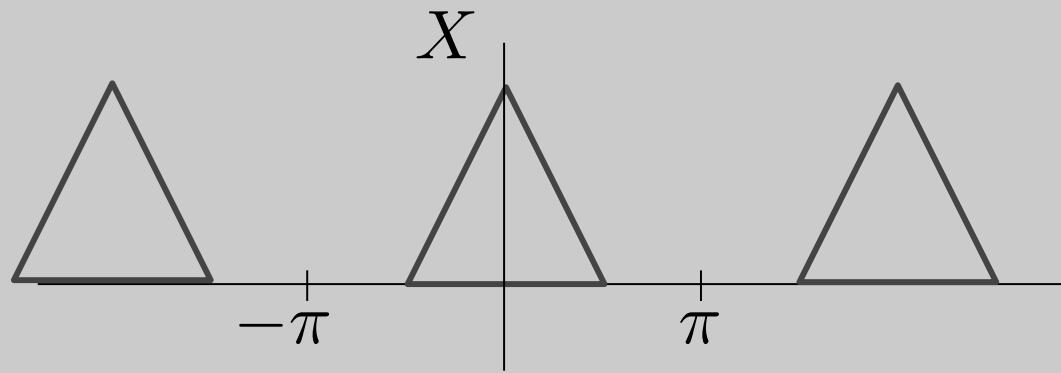


$M=2$

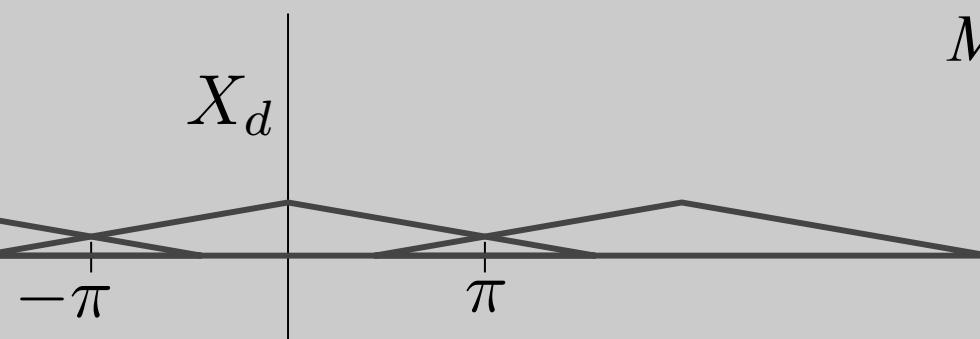


Changing Sampling-rate via D.T Processing

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0} X \left(e^{j(\omega/M - 2\pi i/M)} \right)$$



$M=3$



Anti-Aliasing

