

Lecture 3

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A couple of things

- · Read Ch 2 2.0-2.9
- It's OK to use 2nd edition
- · Class webcasted in bcourses.berkeley.edu
- My office hours: posted on-line
 - W 11a-12 (EE225E priority), Th 3p-5p Cory 506
- Frank Ong
 - Th 5p-6 504 Cory (this week W 2p-3 Cory 400)
 - Lab Bash Tu 2p-3p Cory 521
- · Reward: 1\$ for every typo/errors in my slides/slide

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Ham Stuff

- · Exam:
 - Wednesday, Feb 26th at 6:00 PM, Location TBD
 - Bring your own pencil, pen
 - Bring \$15 or a check for \$15 made out to ARRL-VEC
 - Bring a legal photo ID (passport, driver's license); school ID is not sufficient alone, but must be combined with social security card, birth certificate, or other documents (see http://www.arrl.org/what-to-bring-to-an-exam-session)
 - Apply for a Federal Registration Number (FRN) before the exam, and have that number with you at the exam
 - If you already have a license, bring <u>both</u> original and photocopy of the license (or CSCE) to the exam
 - You could also get licensed on your own:
 http://www.arrl.org/find-an-amateur-radio-license-exam-session

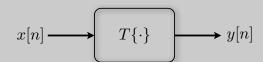
More Ham Stuff

- · Get the book.
- Ham exam preparation lectures
 - -Next 3 weeks
- Wednesday noon, demonstration of satellite communication in Memorial Glade

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Discrete Time Systems

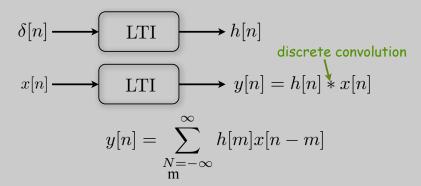


- Causality
- Memoryless
- · Linearity
- Time Invariance
- · BIBO stability

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Discrete-Time LTI Systems

 The impulse response h[n] completely characterizes an LTI system "DNA of LTI"



Sum of weighted, delayed impulse responses!

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BIBO Stability of LTI Systems

 An LTI system is BIBO stable iff h[n] is absolutely summable

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

BIBO Stability of LTI Systems

· Proof: "if"

Proof: If
$$|y[n]| = \left|\sum_{k=-\infty}^{\infty} h[k]x[n-k]\right|$$

$$\leq \sum_{k=-\infty}^{\infty} |h[k]| \cdot |x[n-k]|$$

$$\leq B_x \sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

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7

BIBO Stability of LTI Systems

- Proof: "only if"
 - -suppose $\sum_{k=-\infty}^{\infty}|h[k]|=\infty$ show that there exists bounded x[n] that gives unbounded y[n]

-Let:
$$x[n] = \frac{h[-n]}{|h[-n]|} = \operatorname{Sign}\{h[-n]\}$$

$$y[n] = \sum h[k]x[n-k]$$

$$y[0] = \sum h[k]x[-k] = \sum h[k]h[k]/|h[k]| = \sum |h[k]| = \infty$$

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Discrete-Time Fourier Transform (DTFT)

$$X(e^{j\omega})=\sum_{k=-\infty}^\infty x[k]e^{-j\omega k}$$
 Why one is sum and the other integral?
$$x[n]=rac{1}{2\pi}\int_{-\pi}^\pi X(e^{j\omega})e^{j\omega n}d\omega$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

Why use one over the other?

Alternative

$$X(f) = \sum_{k=-\infty}^{\infty} x[k]e^{-j2\pi fk}$$
$$x[n] = \int_{0.5}^{0.5} X(f)e^{j2\pi fn}df$$

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Example 1:



 $W(e^{j\omega}) = \sum_{k=0}^{N} e^{-j\omega k}$ $= e^{-j\omega N} \left(1 + e^{j\omega} + \dots + e^{j\omega 2N} \right)$

Recall:
$$1 + p + p^2 + \dots + p^M = \frac{1 - p^{M+1}}{1 - p}$$
 $p = e^{j\omega}$ $M = 2N$

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Example 1 cont.

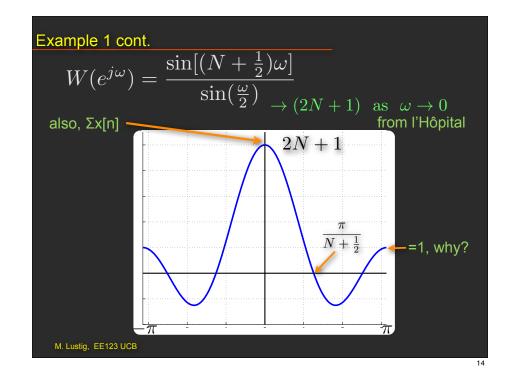
DTFT:

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Example 1 cont.

DTFT:

$$\begin{array}{lll} W(e^{j\omega}) & = & e^{-j\omega N} \left(1 + e^{j\omega} + \cdots + e^{j\omega 2N}\right) \\ & = & e^{-jwN} \frac{1 - e^{j\omega(2N+1)}}{1 - e^{j\omega}} \\ & = & \frac{e^{-jwN} - e^{j\omega N} e^{j\omega}}{1 - e^{j\omega}} & \frac{\times e^{-j\frac{\omega}{2}}}{\times e^{-j\frac{\omega}{2}}} \\ & = & \frac{e^{-j\omega(N+\frac{1}{2})} - e^{j\omega(N+\frac{1}{2})}}{e^{-j\frac{\omega}{2}} - e^{j\frac{\omega}{2}}} \\ & = & \frac{\sin[(N+\frac{1}{2})\omega]}{\sin(\frac{\omega}{2})} & \text{periodic sinc} \end{array}$$



Properties of the DTFT

 $X(e^{j(\omega+2\pi)}) = X(e^{j\omega})$ Periodicity:

Conjugate Symmetry:

$$X^*(e^{j\omega}) = X(e^{-j\omega})$$
 if x[n] is real

$$\mathcal{R}e\left\{X(e^{-j\omega})\right\} = \mathcal{R}e\left\{X(e^{j\omega})\right\}$$
$$\mathcal{I}m\left\{X(e^{-j\omega})\right\} = -\mathcal{I}m\left\{X(e^{j\omega})\right\}$$

Big deal for: MRI, Communications, more....

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