

EE 120, GSI: Ming

OH: Tu 2-3pm, Cory 406
Jinming@berkeley.edu

HW due Friday (9/2)

- Ice-breaking
- DFT basics
- Phasor analysis

time domain

freq. domain

$$x[n] \longleftrightarrow X[k]$$

$$X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}kn}$$

$$x[n] = \sum_{k=0}^{N-1} X[k] e^{j\frac{2\pi}{N}kn}$$

Why DFT?

- Signal processing: low-pass filter
- communication: AM
- image compression

(a) $x[n] = \cos\left(\frac{2\pi k}{N}n\right)$

$$= \frac{1}{2} e^{j\frac{2\pi k}{N}n} + \frac{1}{2} e^{-j\frac{2\pi k}{N}n}$$

$$X[k] = \frac{1}{N} \sum_{n=0}^{N-1} \left(\frac{1}{2} e^{j\frac{2\pi k}{N}n} + \frac{1}{2} e^{-j\frac{2\pi k}{N}n} \right) e^{-j\frac{2\pi k}{N}kn}$$

$$= \frac{1}{2N} \sum_{n=0}^{N-1} e^{j\frac{2\pi(k-k)n}{N}} + \frac{1}{2N} \sum_{n=0}^{N-1} e^{-j\frac{2\pi(k+k)n}{N}}$$

$$= \begin{cases} \frac{1}{2} & k=K \text{ or } N-K \\ 0 & \text{o.w.} \end{cases}$$

$$\sum_{n=0}^{N-1} e^{j\frac{2\pi kn}{N}} = \frac{1 - e^{j\frac{2\pi kN}{N}}}{1 - e^{j\frac{2\pi k}{N}}} = 0 \text{ if } k \neq 0$$

(b) Signal superposition:

$$\frac{1}{N} \sum_{n=0}^{N-1} (\alpha x[n] + \beta y[n]) e^{-j\frac{2\pi kn}{N}}$$

$$= \frac{1}{N} \sum_{n=0}^{N-1} \alpha x[n] e^{-j\frac{2\pi kn}{N}} + \frac{1}{N} \sum_{n=0}^{N-1} \beta y[n] e^{-j\frac{2\pi kn}{N}}$$

$$= \alpha X[k] + \beta Y[k]$$

(c) $\frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi nk}{N}}$

$$= \frac{1}{N} \sum_{n=0}^{N-1} x[n-M] e^{-j\frac{2\pi(n-M)k}{N}} e^{-j\frac{2\pi Mk}{N}}$$

$$= \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{j\frac{2\pi nk}{N}} e^{-j\frac{2\pi Mk}{N}} = X[k] e^{-j\frac{2\pi Mk}{N}}$$

Note: $x[n]$ periodic

(d) $\frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{j\frac{2\pi nM}{N}} e^{-j\frac{2\pi nk}{N}}$

$$= \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi n(k-M)}{N}} = X[k-M]$$

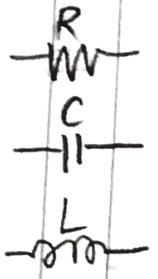
(e) $\frac{1}{N} \sum_{n=0}^{N-1} x[n] y^*[n] = \frac{1}{N} \sum_{n=0}^{N-1} \sum_{k_1=0}^{N-1} \sum_{k_2=0}^{N-1} X[k_1] e^{j\frac{2\pi k_1 n}{N}} Y^*[k_2] e^{-j\frac{2\pi k_2 n}{N}}$

$$= \sum_{k=0}^{N-1} X[k] Y^*[k]$$

$\begin{cases} 0 & \text{if } k_1 \neq k_2 \\ NX[k]Y^*[k] & \text{if } k_1 = k_2 \end{cases}$

2. phasor analysis

time domain



$v(t) = i(t) \cdot R$

$i(t) = C \frac{dv(t)}{dt}$

$v(t) = L \cdot \frac{di(t)}{dt}$

phasor domain

$V(\omega) = I(\omega) \cdot R$

$V(\omega) = I(\omega) \frac{1}{j\omega C}$

$V(\omega) = I(\omega) \cdot j\omega L$

impedance ($\frac{V(\omega)}{I(\omega)}$)

R

$\frac{1}{j\omega C}$

$j\omega L$

Derivation of phasor:

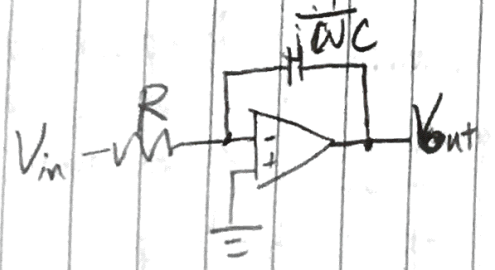
$v_c(t) = \text{Re}\{V_c e^{j\omega t}\}$

$= V_c \cos \omega t$

$i_c(t) = -C \cdot V_c \omega \sin \omega t$

$= \text{Re}\{jC V_c \omega e^{j\omega t}\} \Rightarrow I_c = jC V_c \omega$

$\Rightarrow Z_c = \frac{V_c}{I_c} = \frac{V_c}{jC V_c \omega} = \frac{1}{j\omega C}$



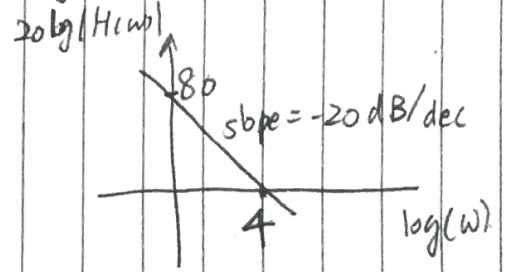
$V_- = V_+ = 0$

$\frac{V_{out}}{\frac{1}{j\omega C}} = -\frac{V_{in}}{R}$

$H(\omega) = \frac{V_{out}}{V_{in}} = -\frac{1}{j\omega CR}$

$H(\omega) = -\frac{10^4}{j\omega}$

$\log |H(\omega)| = \log \frac{10^4}{\omega} = 4 - \log \omega$



(c).

KCL:

$-\frac{V_{in}}{R} = C \cdot \frac{dV_{out}}{dt}$

$V_{out} = -\frac{1}{RC} \int V_{in} dt$
integrates V_{in}