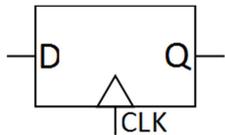


## State Elements

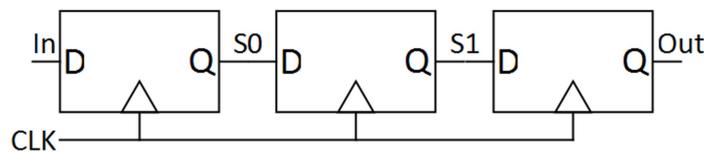
State elements provide a means of storing values, and controlling the flow of information in the circuit. The most basic state element (we're concerned with) is a DQ Flip-Flop:



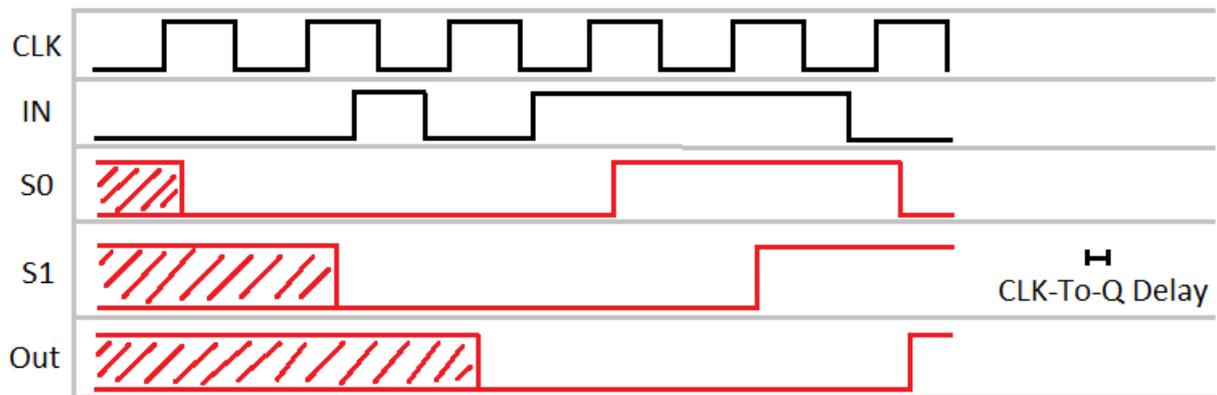
D is a single bit input, Q is a single bit output. On the **rising edge** of the clock, the value from D is copied to Q, after a small delay (known as the CLK-to-Q delay). At all other times, Q presents the value last copied and ignores D.

An n-bit register is just n DQ Flip-Flops aligned in parallel, tied to the same clock.

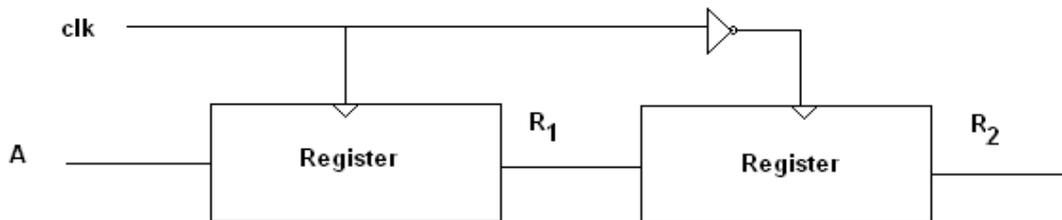
**Timing Diagram 1:** Fill out the timing diagram for the circuit below



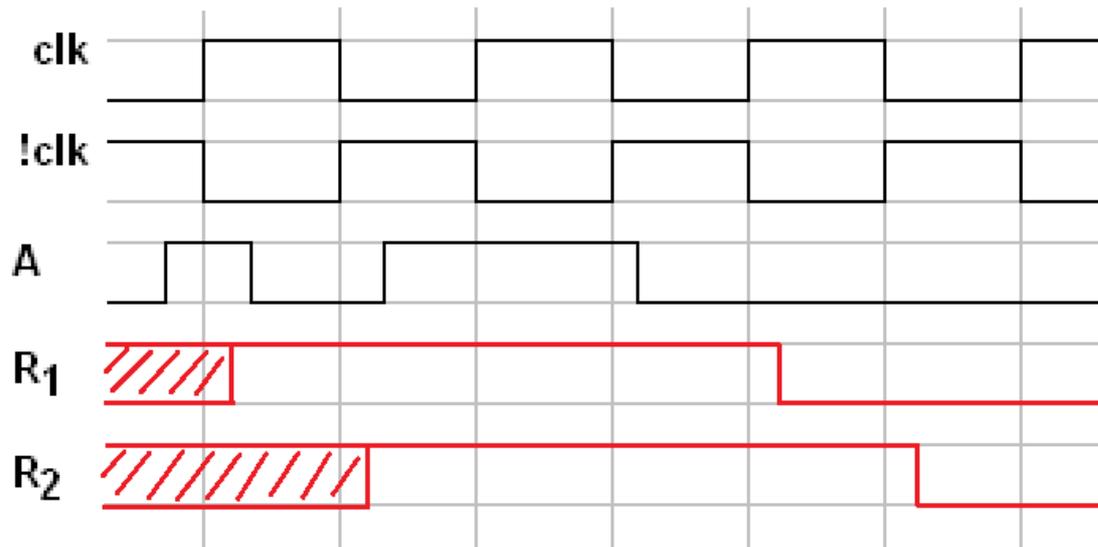
**Note:** An unknown value is "shaded."



**Timing Diagram 2:** Fill out the timing diagram for the circuit below

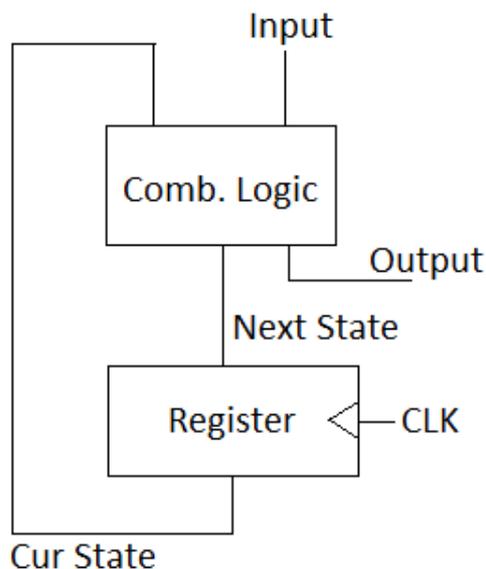


**Note:** An unknown value is "shaded."



### Finite State Machines

FSMs can be an incredibly useful computational tool. They have a straightforward implementation in hardware:

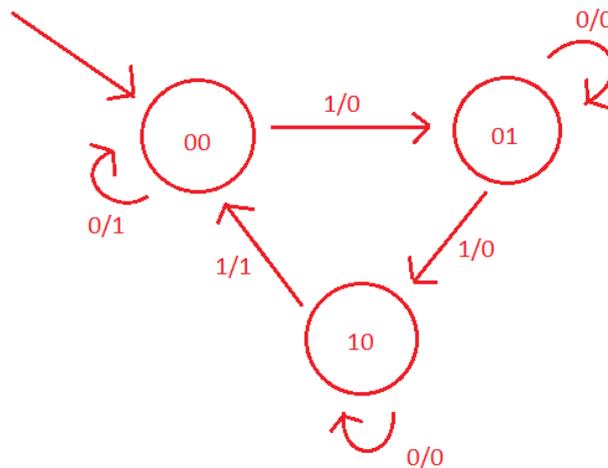


The register holds the current state (encoded as a particular combination of bits), and the combinational logic block maps from {current state, input} to {next state, output}.

### Exercises

Draw a transition diagram for an FSM that can take in an input sequence one bit at a time, and after each input is received, output whether the number of 1s is divisible by 3. Write out the truth table that the combinational logic block must implement (remember to assign each state a binary encoding). Finally, write the Boolean algebra expressions that implement the FSM's truth table.

## CS 61C Summer 2012 Discussion 11 – State, Timing, and CPU (Solutions)



The states each correspond to the number of 1s seen so far, mod 3. When this quantity is 0, 1s seen so far is divisible by 3, and we output 1.

We assign our three states the encodings 00, 01, 10. Note that these encodings are completely arbitrary – but other encodings (i.e. 01, 10, 11) will lead to a different truth table and final combinational logic.

cur. state	input	next state	output
00	0	00	1
00	1	01	0
01	0	01	0
01	1	10	0
10	0	10	0
10	1	00	1
11	0	XX	X
11	1	XX	X

Behavior for current state 11 is “undefined” since we don’t expect our machine to ever reach that state. We represent this in the truth table above using an ‘x’ to stand for “don’t care.” We can use this to our advantage by choosing values (0 or 1 for each x) that will simplify our combinational logic.

Boolean logic expressions (CS[1],CS[0] – bits of current state, I – input, NS[1], NS[0] – next state, Out – output). Overbar means complement.

NS[1]:

Direct method:  $\overline{CS[1]} \cdot CS[0] \cdot I + CS[1] \cdot CS[0] \cdot \bar{I}$

Simplest possible:  $CS[0] \cdot I + CS[1] \cdot \bar{I}$

(Simplest version assumes  $NS[1] = 1$  for both  $CS[1] \cdot CS[0]$ )

NS[0]:

## CS 61C Summer 2012 Discussion 11 – State, Timing, and CPU (Solutions)

---

Direct method:  $\overline{CS[1]} \cdot \overline{CS[0]} \cdot I + \overline{CS[1]} \cdot CS[0] \cdot \bar{I}$

Simplest possible:  $\overline{CS[1]} \cdot \overline{CS[0]} \cdot I + CS[0] \cdot \bar{I}$

(Simplest version assumes  $NS[0] = 1$  for  $CS[1] \cdot CS[0] \cdot \bar{I}$  and  $NS[0] = 0$  for  $CS[1] \cdot CS[0] \cdot I$ )

Out:

Direct method:  $\overline{CS[1]} \cdot \overline{CS[0]} \cdot \bar{I} + CS[1] \cdot \overline{CS[0]} \cdot I$

Simplest possible:  $\overline{CS[1]} \cdot \overline{CS[0]} \cdot \bar{I} + CS[1] \cdot I$

(Simplest version assumes  $Out = 0$  for  $CS[1] \cdot CS[0] \cdot \bar{I}$  and  $Out = 1$  for  $CS[1] \cdot CS[0] \cdot I$ )

## CS 61C Summer 2012 Discussion 11 – State, Timing, and CPU (Solutions)

---

### Clocking Methodology

- The input signal to each state element must stabilize before each rising edge.
- Critical path: Longest delay path between state elements in the circuit.
- Min clock period =  $t_{\text{clk-to-q}} + t_{\text{CL}} + t_{\text{setup}}$ , where  $t_{\text{CL}}$  is the Combinational Logic delay in the critical path.
- If we place registers in the critical path, we can shorten the period by reducing the amount of logic between registers

### Clocking Problem

- The circuit below computes the weighted average of 4 values
- Logic Delays -  $t_{\text{mult}} = 55\text{ns}$ ,  $t_{\text{add}} = 19\text{ns}$ ,  $t_{\text{shift}} = 2\text{ns}$
- Register Parameters -  $t_{\text{setup}} = 2\text{ns}$ ,  $t_{\text{hold}} = 1\text{ns}$ ,  $t_{\text{clk-to-q}} = 3\text{ns}$

What is the critical path delay and the maximum clock rate this circuit can operate at?

Critical path – path from a top register to the bottom register.

$\text{Clk-to-q} + \text{mult} + \text{add} + \text{add} + \text{shift} + \text{setup} = 3 + 55 + 19 + 19 + 2 + 2 = 100 \text{ ns.}$

Max frequency =  $1/\text{Min period} = 10 \text{ MHz}$

If you add one stage of registers (pipelining), what is the highest clock rate you can get?

Best to add registers in a place that minimizes the longest path through the circuit. This would be right after the multiplication.

Path from top registers to middle registers is  $\text{clk-to-q} + \text{mult} + \text{setup} = 3 + 55 + 2 = 60 \text{ ns.}$

Path from middle to bottom registers is  $\text{clk-to-q} + \text{add} + \text{add} + \text{shift} + \text{setup} = 3 + 19 + 19 + 2 + 2 = 45 \text{ ns.}$

So our new critical path has Min period of 60 ns, so our Max frequency = 16.67 MHz.

CS 61C Summer 2012 Discussion 11 - State, Timing, and CPU (Solutions)

---

