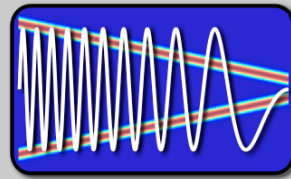


EE123



Digital Signal Processing

Lecture 19

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Topics

- Last time
 - Upsampling
 - Resampling by rational fraction
- Today
 - Interchanging Compressors/Expanders with filtering
 - Polyphase decomposition
 - Multi-rate processing

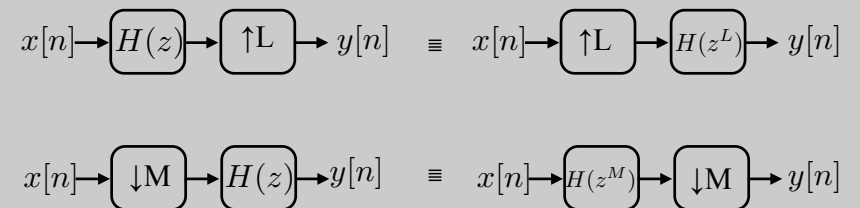
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Multi-Rate Signal Processing

- What if we want to resample by 1.01T?
 - Expand by $L=100$
 - Filter $\pi/101$ (\$\$\$\$\$)
 - Downsample by $M=101$
- Fortunately there are ways around it!
 - Called multi-rate
 - Uses compressors, expanders and filtering

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Interchanging Operations

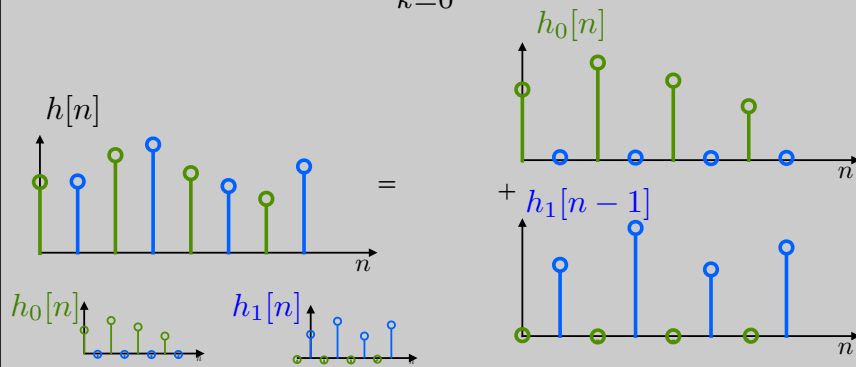


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Polyphase Decomposition

- We can decomposed an impulse response to:

$$h[n] = \sum_{k=0}^{M-1} h_k[n - k]$$



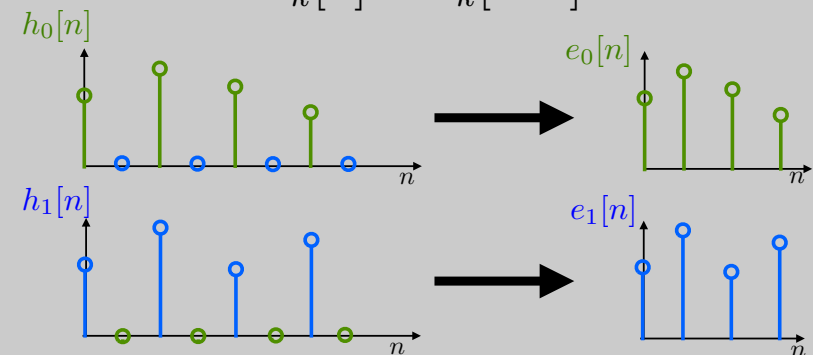
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Polyphase Decomposition

- Define:

$$h_k[n] \rightarrow \downarrow M \rightarrow e_k[n]$$

$$e_k[n] = h_k[nM]$$



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Polyphase Decomposition

$$e_k[n] \rightarrow \uparrow M \rightarrow h_k[n]$$

recall upsampling \Rightarrow scaling

$$H_k(z) = E_k(z^M)$$

Also, recall:

$$h[n] = \sum_{k=0}^{M-1} h_k[n - k]$$

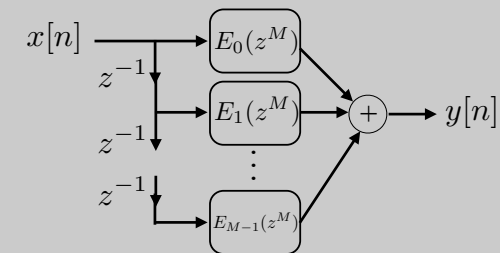
So,

$$H(z) = \sum_{k=0}^{M-1} E_k(z^M) z^{-k}$$

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Polyphase Decomposition

$$H(z) = \sum_{k=0}^{M-1} E_k(z^M) z^{-k}$$



Why should you care?

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Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

• Problem:

–Compute all $y[n]$ and then throw away --
wasted computation!

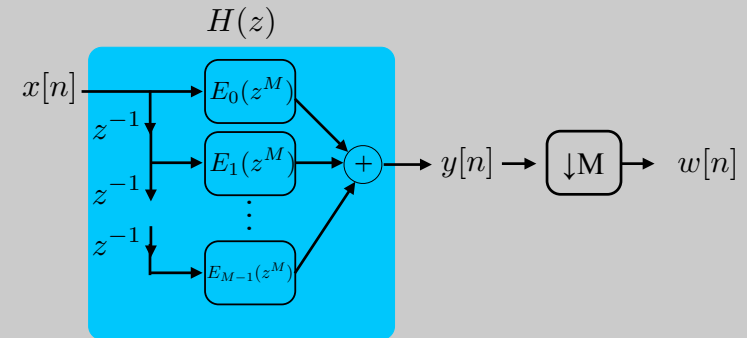
- For FIR length $N \Rightarrow N$ mults/unit time

–Can interchange Filter with compressor?

- Not in general!

Polyphase Implementation of Decimation

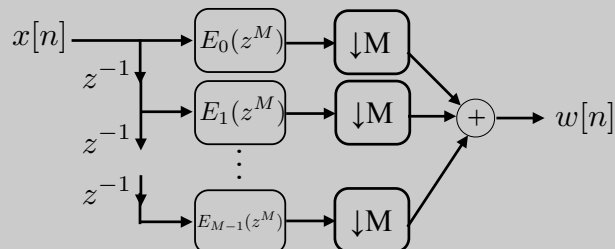
$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$



Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

Interchange filter with decimation

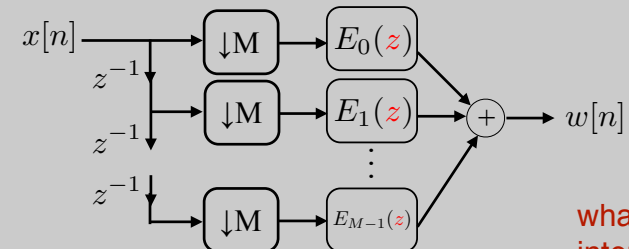


now, what can we do?

Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

Interchange filter with decimation



what about
interpolation?

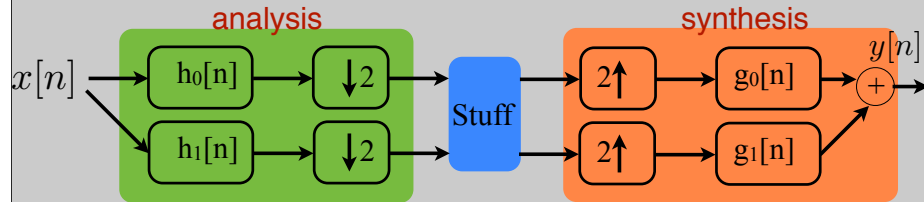
Computation:

Each Filter: $N/M * (1/M)$ mult/unit time

Total: N/M mult/unit time

Multirate FilterBank

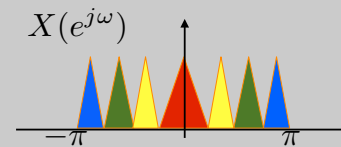
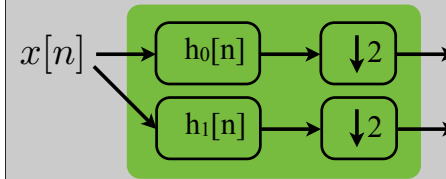
- $h_0[n]$ is low-pass, $h_1[n]$ is high-pass
- Often $h_1[n] = e^{j\pi n} h_0[n]$ or $H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$



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Subtleties in Time-Freq Tiling

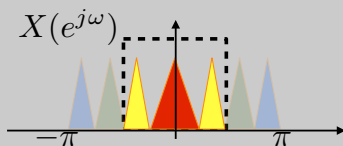
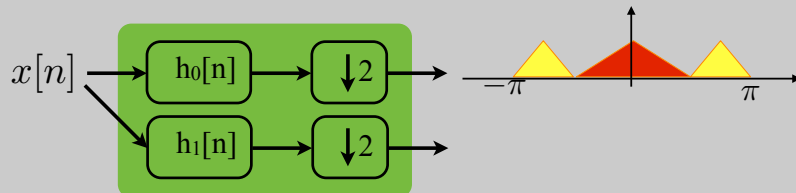
- Assume h_0, h_1 are ideal low,high pass filters



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Subtleties in Time-Freq Tiling

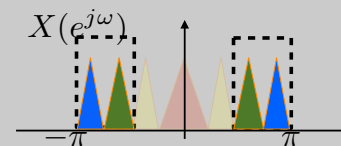
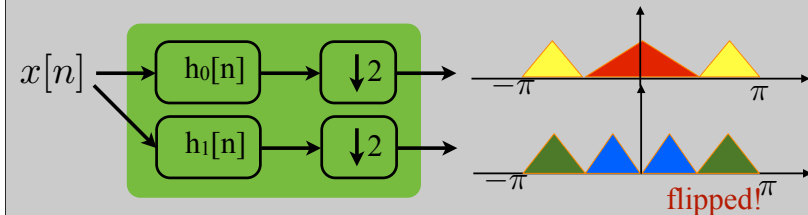
- Assume h_0, h_1 are ideal low,high pass filters



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Subtleties in Time-Freq Tiling

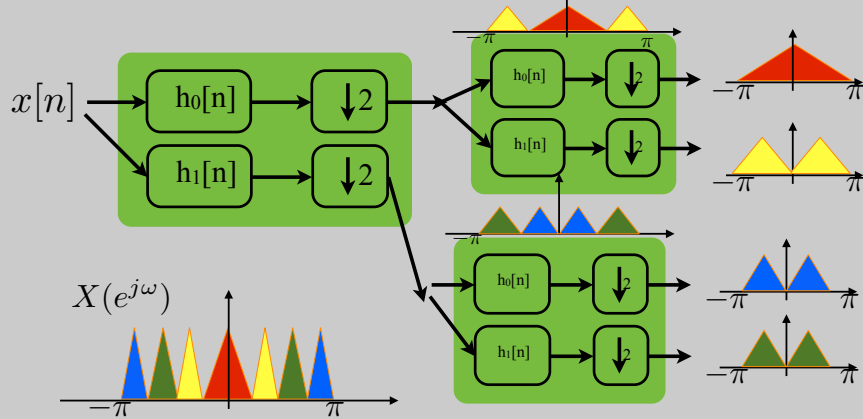
- Assume h_0, h_1 are ideal low,high pass filters



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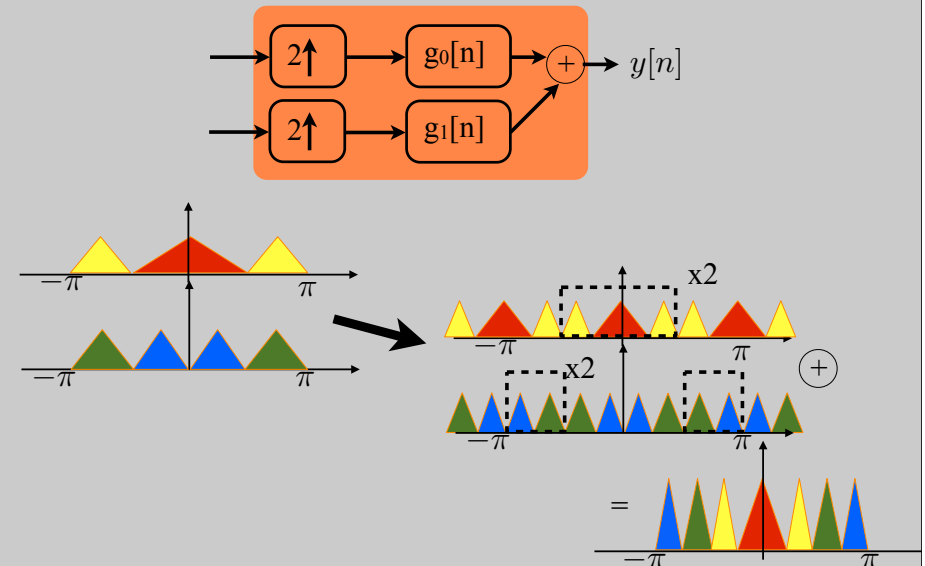
Subtleties in Time-Freq Tiling

- Assume h_0, h_1 are ideal low, high pass filters



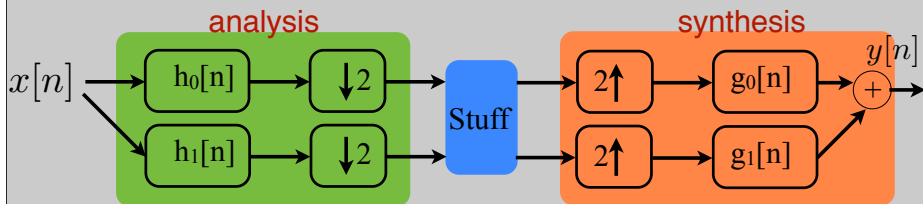
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Perfect Reconstruction Ideal Filters



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Perfect Reconstruction non-Ideal Filters



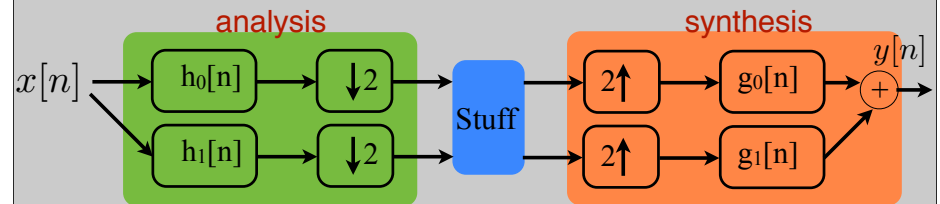
$$Y(e^{j\omega}) = \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j\omega}) + G_1(e^{j\omega})H_1(e^{j\omega})] X(e^{j\omega}) + \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j(\omega-\pi)}) + G_1(e^{j\omega})H_1(e^{j(\omega-\pi)})] X(e^{j(\omega-\pi)})$$

need to cancel!

aliasing

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Quadrature Mirror Filters - perfect recon



QMF - mirror around $\pi/2$

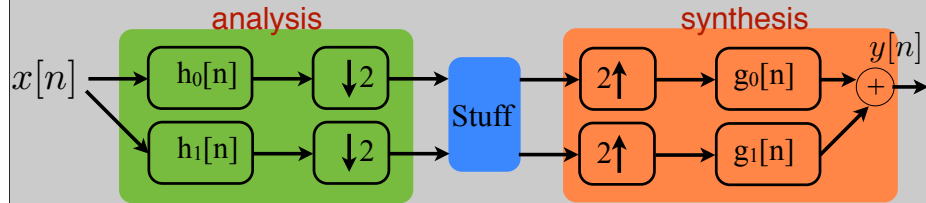
$$H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$$

$$G_0(e^{j\omega}) = 2H_0(e^{j\omega})$$

$$G_1(e^{j\omega}) = -2H_1(e^{j\omega})$$

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Quadrature Mirror Filters - perfect recon



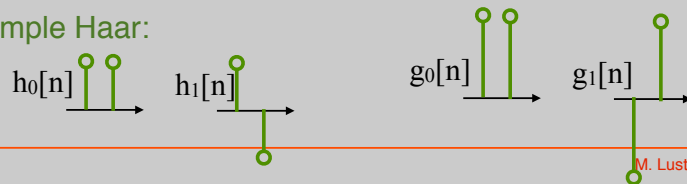
QMF - mirror around $\pi/2$

$$H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$$

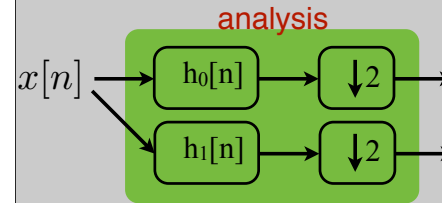
$$G_0(e^{j\omega}) = 2H_0(e^{j\omega})$$

$$G_1(e^{j\omega}) = -2H_1(e^{j\omega})$$

Example Haar:



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$$e_{00} = h_0[2n]$$

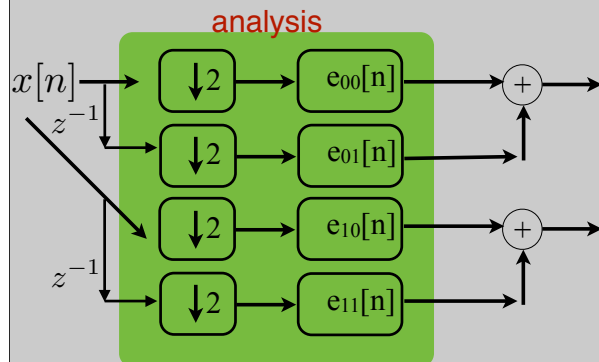
$$e_{01} = h_0[2n + 1]$$

$$e_{10} = h_1[2n] = e^{j2\pi n} h_0[2n] = e_{00}[n]$$

$$e_{11} = h_1[2n + 1] = e^{j2\pi n} e^{j\pi} h_0[2n + 1] = -e_{01}[n]$$

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Polyphase Filter-Bank



$$e_{00} = h_0[2n]$$

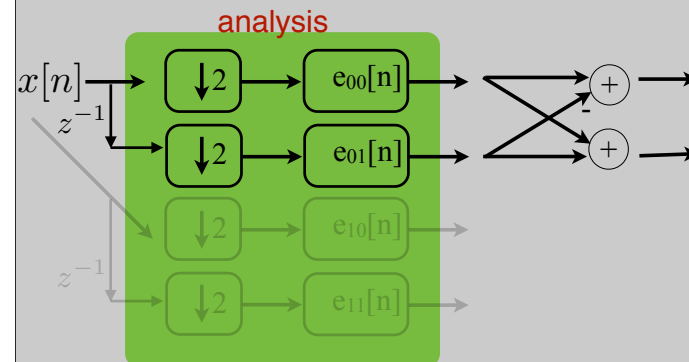
$$e_{01} = h_0[2n + 1]$$

$$e_{10} = e_{00}[n]$$

$$e_{11} = -e_{01}[n]$$

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Polyphase Filter-Bank



$$e_{00} = h_0[2n]$$

$$e_{01} = h_0[2n + 1]$$

$$e_{10} = e_{00}[n]$$

$$e_{11} = -e_{01}[n]$$

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