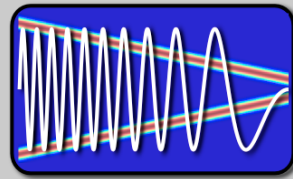


EE123



Digital Signal Processing

Lecture 17

M. Lustig, EECS UC Berkeley

Carl Sagan - Cosmos



Billions

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Topics

- Last time
 - Changing Sampling Rate via DSP
 - Downsampling
- Today
 - Upsampling

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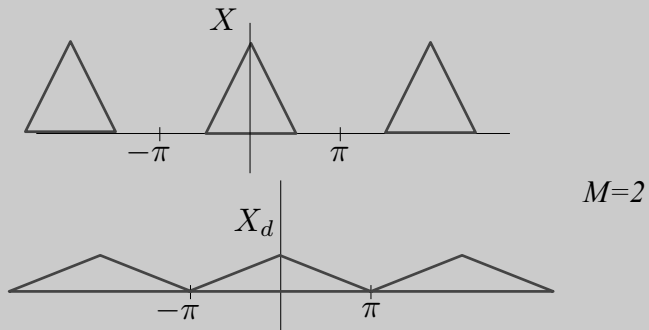
DownSampling

- Much like C/D conversion
- Expect similar effects:
 - Aliasing
 - mitigate by antialiasing filter
- Finely sampled signal \Rightarrow almost continuous
 - Downsample in that case is like sampling!

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Changing Sampling-rate via D.T Processing

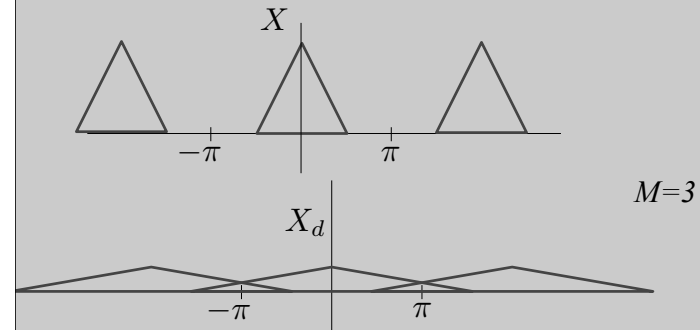
$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(e^{j(\omega/M - 2\pi i/M)} \right)$$



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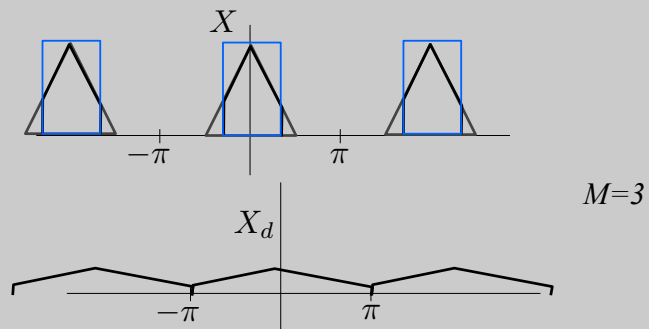
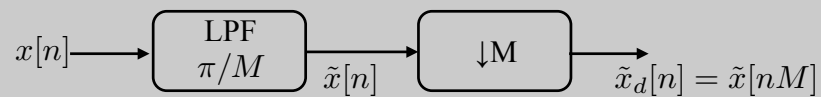
Changing Sampling-rate via D.T Processing

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X \left(e^{j(\omega/M - 2\pi i/M)} \right)$$



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Anti-Aliasing



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UpSampling

- Much like D/C converter
- Upsample by A LOT \Rightarrow almost continuous
- Intuition:
 - Recall our D/C model: $x[n] \Rightarrow x_s(t) \Rightarrow x_c(t)$
 - Approximate “ $x_s(t)$ ” by placing zeros between samples
 - Convolve with a sinc to obtain “ $x_c(t)$ ”

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Up-sampling

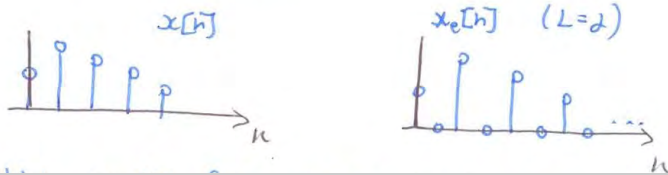
$$x[n] = X_c(nT)$$

$$x_e[n] = X_c(nT') \quad \text{where } T' = \frac{T}{L}, \quad L \text{ integer}$$

obtain $x_i[n]$ from $x_e[n]$ in two steps:

1) Generate $x_e[n] = \begin{cases} x[n/L] & n = 0, \pm L, \pm 2L, \dots \\ 0 & \text{otherwise} \end{cases}$

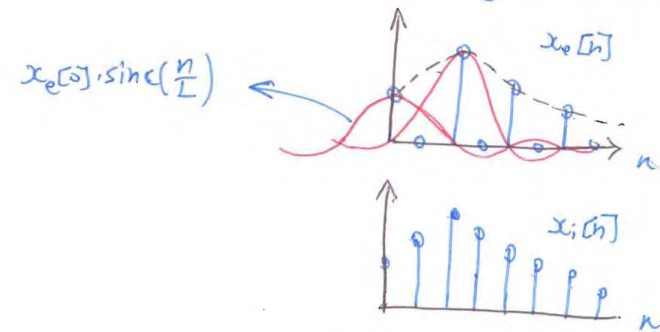
↓ expansion



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Up-Sampling

2) Obtain $x_i[n]$ from $x_e[n]$ by bandlimited interpolation.



$$x_i[n] = x_e[n] * \text{sinc}\left(\frac{n}{L}\right)$$

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Up-Sampling

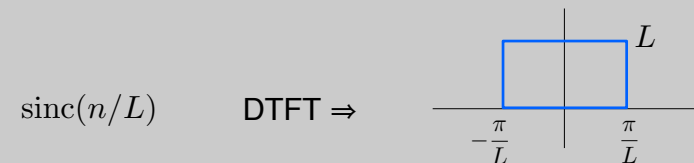
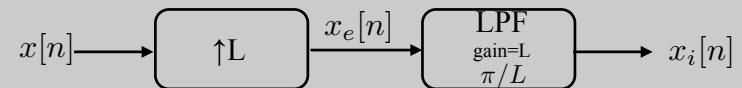
$$x_i[n] = x_e[n] * \text{sinc}(n/L)$$

$$x_e[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - kL]$$

$$x_i[n] = \sum_{k=-\infty}^{\infty} x[k] \text{sinc}\left(\frac{n - kL}{L}\right)$$

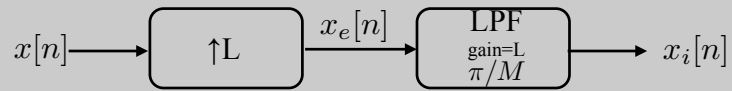
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Frequency Domain Interpretation



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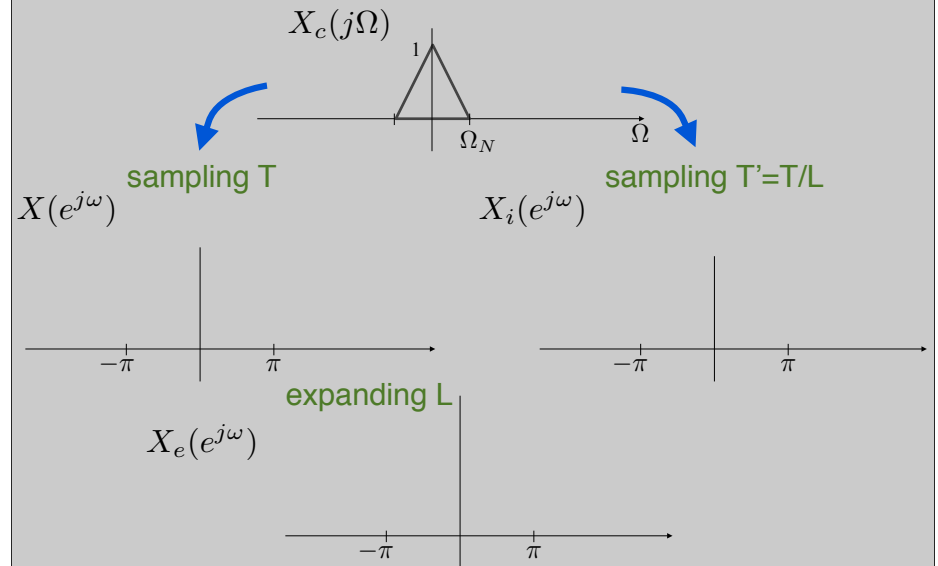
Frequency Domain Interpretation



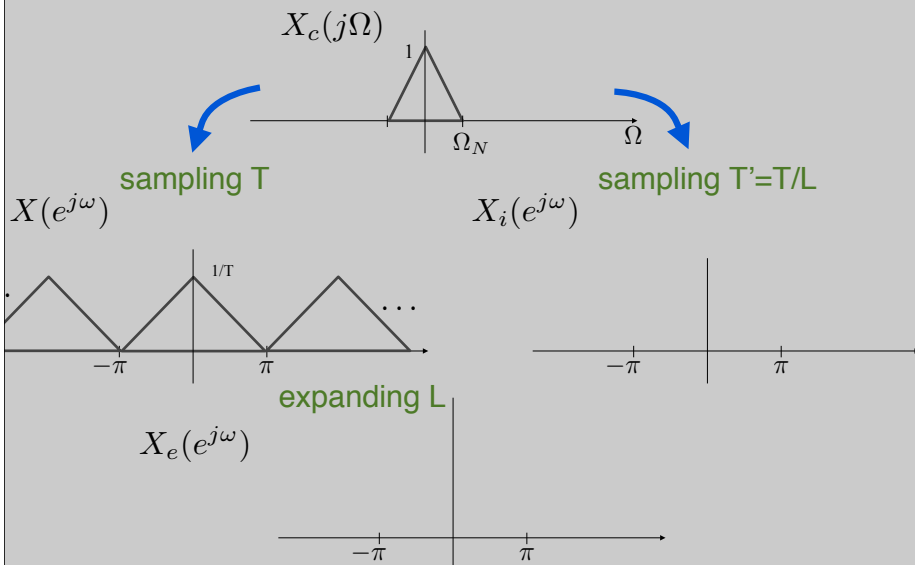
$$\begin{aligned}
 X_e(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} \underbrace{x_e[n]}_{\neq 0 \text{ only for } n=mL \text{ (integer } m)} e^{-j\omega n} \\
 &= \sum_{m=-\infty}^{\infty} \underbrace{x_e[mL]}_{=x[m]} e^{-j\omega mL} = X(e^{j\omega L})
 \end{aligned}$$

Compress DTFT by a factor of L!

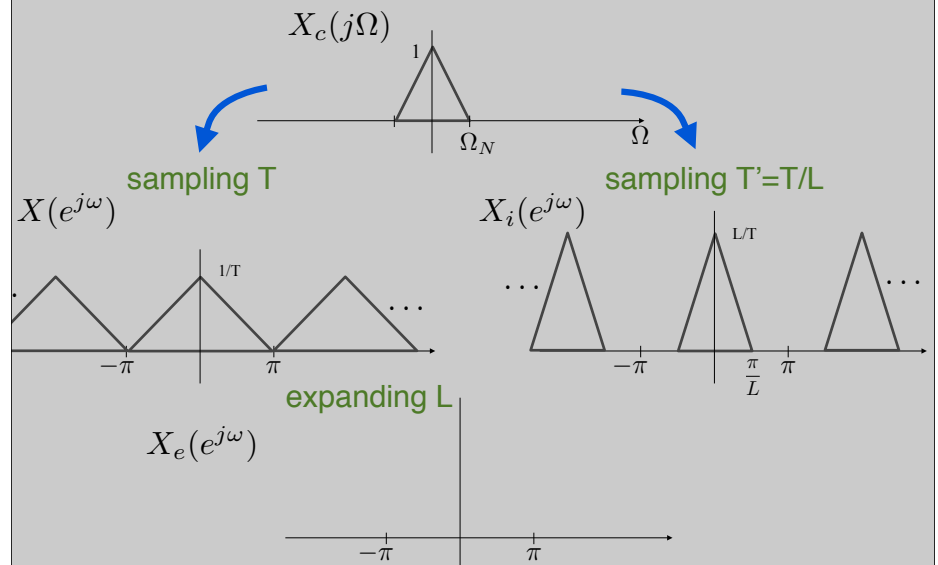
Example:



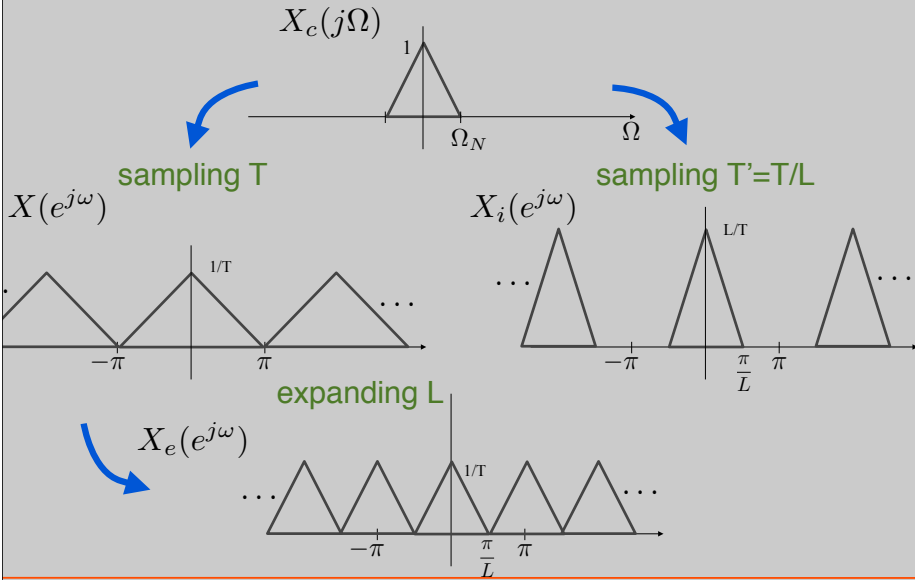
Example:



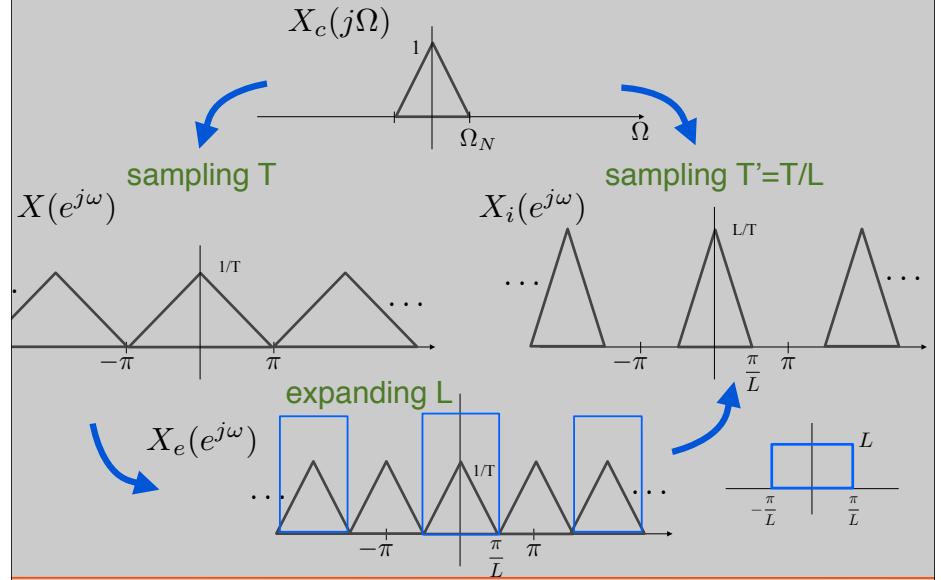
Example:



Example:

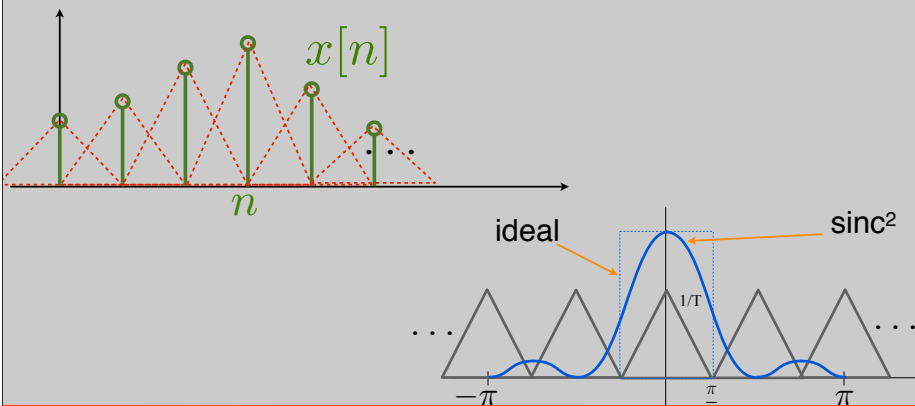


Example:



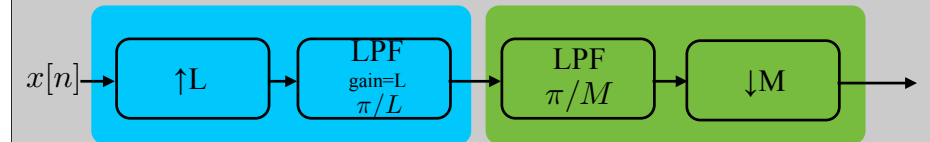
Practical Upsampling

- Can interpolate with simple, practical filters. What's happening?
- Example: $L=3$, linear interpolation - convolve with triangle

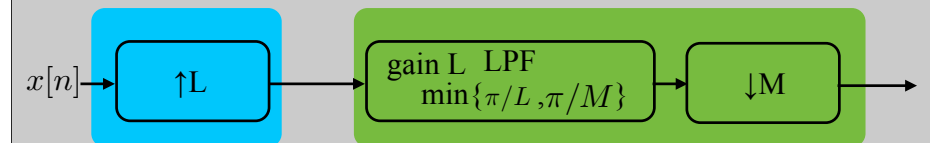


Resampling by non-integer

- $T' = TM/L$ (upsample L, downsample M)



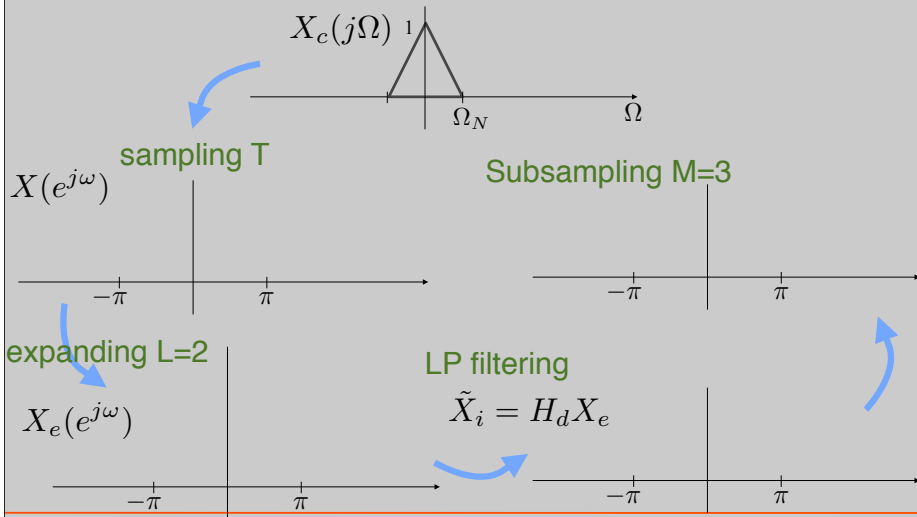
Or,



- What would happen if change order?

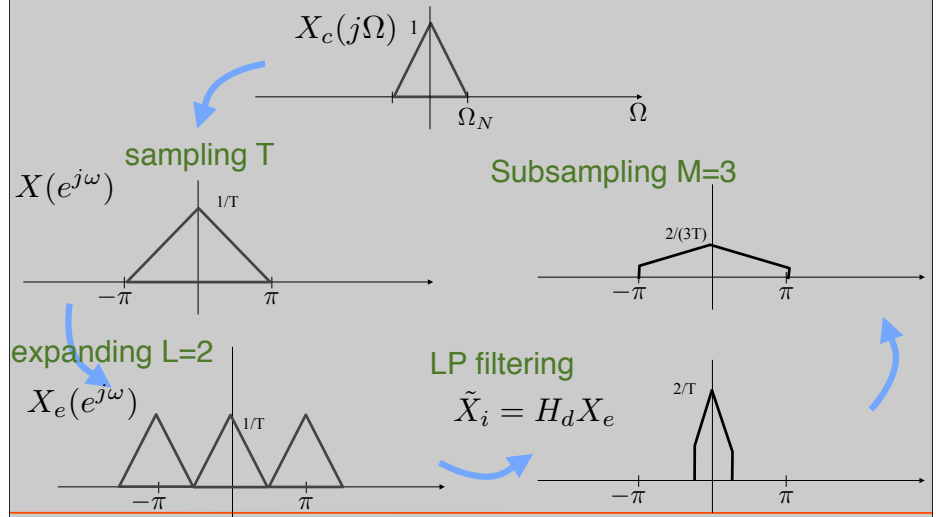
Example:

- $L = 2, M=3, T'=3/2T$ (fig 4.30)



Example:

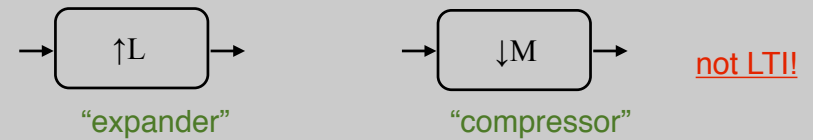
- $L = 2, M=3, T'=3/2T$ (fig 4.30)



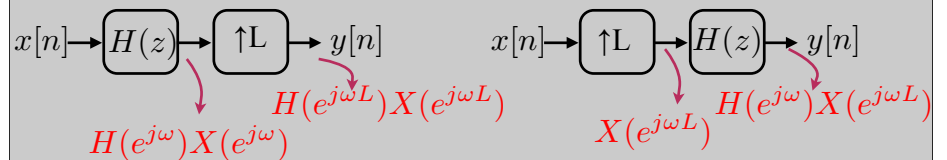
Multi-Rate Signal Processing

- What if we want to resample by $1.01T$?
 - Expand by $L=100$
 - Filter $\pi/101$ (\$\$\$\$\$)
 - Downsample by $M=101$
- Fortunately there are ways around it!
 - Called multi-rate
 - Uses compressors, expanders and filtering

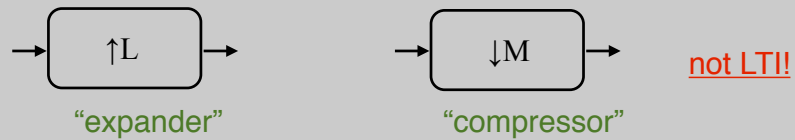
Interchanging Operations



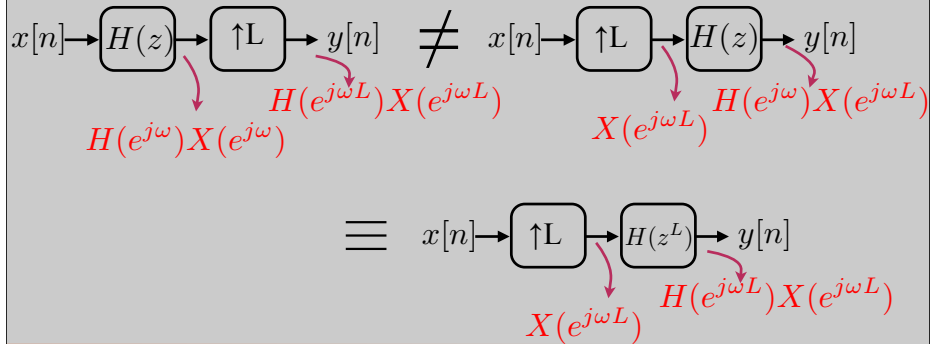
Note:



Interchanging Operations



Note:



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Interchanging Filter Expander

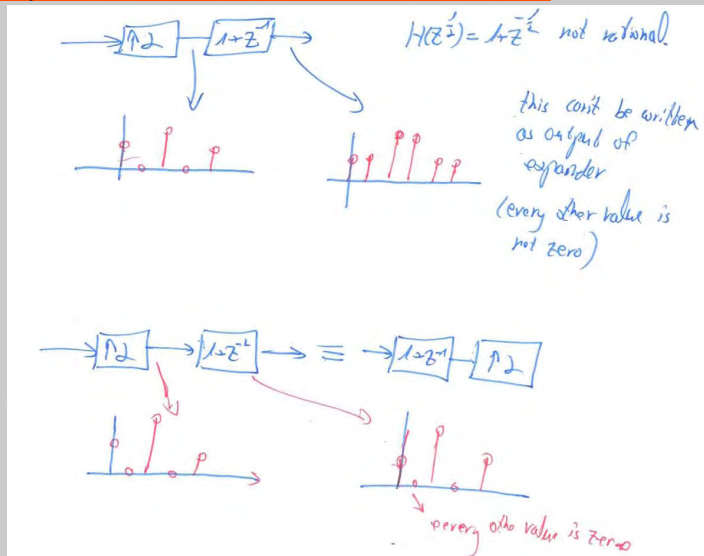
- Q: Can we move expander from Left to Right (with xform)?



- A: Yes, if $H(z)$ is rational
No, otherwise

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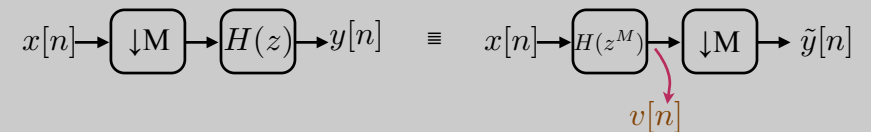
Example:



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Compressor

Claim:



Proof:

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Compressor

Proof:

$$\begin{aligned}
 Y(e^{j\omega}) &= H(e^{j\omega}) \left(\frac{1}{M} \sum_{i=0}^{M-1} x(e^{j(\frac{\omega}{M} - \frac{2\pi i}{M})}) \right) = \\
 &= \frac{1}{M} \sum_{i=0}^{M-1} \underbrace{H(e^{j(\omega - 2\pi i)})}_{= H(e^{j\omega})} X(e^{j(\frac{\omega}{M} - \frac{2\pi i}{M})}) \\
 &= \frac{1}{M} \sum_{i=0}^{M-1} H(e^{jM(\frac{\omega}{M} - \frac{2\pi i}{M})}) X(e^{j(\frac{\omega}{M} - \frac{2\pi i}{M})})
 \end{aligned}$$

$$V(e^{j\omega}) = H(e^{j\omega M}) X(e^{j\omega})$$

after compressor

Compressor

Claim:



Proof:

$$\begin{aligned}
 Y(e^{j\omega}) &= H(e^{j\omega}) \left(\frac{1}{M} \sum_{i=0}^{M-1} x(e^{j(\frac{\omega}{M} - \frac{2\pi i}{M})}) \right) = \\
 &= \frac{1}{M} \sum_{i=0}^{M-1} \underbrace{H(e^{j(\omega - 2\pi i)})}_{= H(e^{j\omega})} X(e^{j(\frac{\omega}{M} - \frac{2\pi i}{M})}) \\
 &= \frac{1}{M} \sum_{i=0}^{M-1} H(e^{jM(\frac{\omega}{M} - \frac{2\pi i}{M})}) X(e^{j(\frac{\omega}{M} - \frac{2\pi i}{M})})
 \end{aligned}$$

$$V(e^{j\omega}) = H(e^{j\omega M}) X(e^{j\omega})$$

v[n]

after compressor

Q: How compress from right to left?
 A: only if $H(z^M)$ rational.

Multi-Rate Filtering



Narrow band
 very sharp filter \rightarrow long impulse response.

faster to do 2 stage.