

EE120 - Fall'15 - Lecture 16 Notes¹

Murat Arcak

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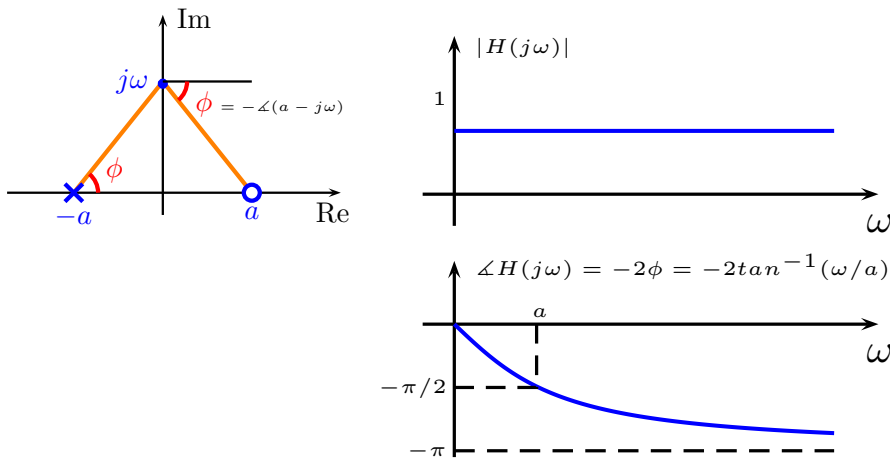
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All-Pass Systems

Section 9.4.3 in Oppenheim & Willsky

What is the frequency response of an LTI system with transfer function:

$$H(s) = \frac{a - s}{s + a}, \quad a > 0?$$



General all-pass system:

$$H_{ap}(s) = \frac{\prod_{i=1}^n (a_i - s)}{\prod_{i=1}^n (s + a_i)}, \quad a_i > 0 \quad i = 1, \dots, n. \quad (1)$$

For a stable and causal all-pass system, all zeros are in the right half-plane because they are mirror images of the poles.

Although $|H_{ap}(j\omega)| \equiv 1$, an all-pass system introduces delay:

$$e^{j\omega t} \rightarrow \boxed{H_{ap}(j\omega)} \rightarrow H_{ap}(j\omega)e^{j\omega t} = e^{j\angle H_{ap}(j\omega)} e^{j\omega t} = e^{j\omega(t - \tau(\omega))}$$

where:

$$\tau(\omega) \triangleq -\frac{\angle H_{ap}(j\omega)}{\omega} > 0.$$

Moreover, the system is not linear phase (i.e. $\tau(\omega)$ is not constant); therefore it causes phase distortion. (Recall Lecture 8.)

Minimum Phase Systems

A stable and causal LTI system is called *minimum phase* if all of its zeros are in the open left half-plane.

Any non-minimum phase transfer function $H(s)$ can be decomposed as:

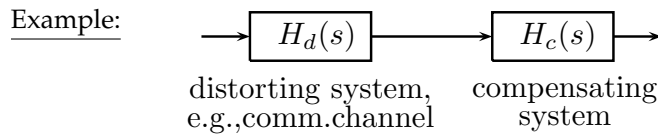
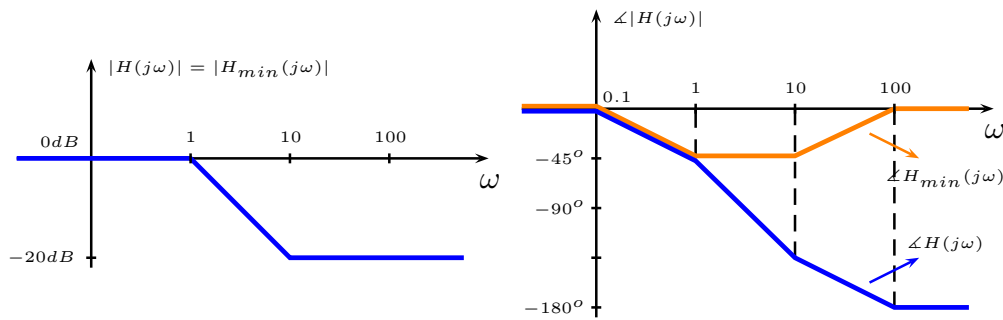
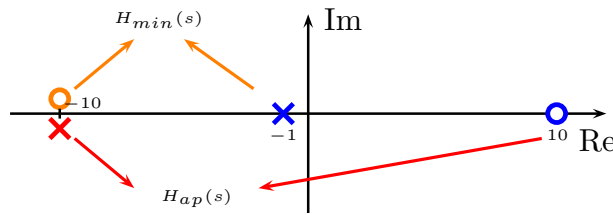
$$H(s) = \underbrace{H_{min}(s)}_{\text{min-phase}} \underbrace{H_{ap}(s)}_{\text{all-pass}} \quad (2)$$

This decomposition explains the genesis of the term *minimum phase*:

$$|H(j\omega)| = |H_{min}(j\omega)| \text{ since } |H_{ap}(j\omega)| \equiv 1,$$

but the all-pass component adds more delay. Therefore, $H(s)$ and $H_{min}(s)$ have identical frequency responses in magnitude, but $H_{min}(s)$ has the minimum phase delay.

Example:
$$H(s) = \frac{10 - s}{10(s + 1)} = \underbrace{\frac{s + 10}{10(s + 1)}}_{H_{min}(s)} \cdot \underbrace{\frac{10 - s}{s + 10}}_{H_{ap}(s)}$$

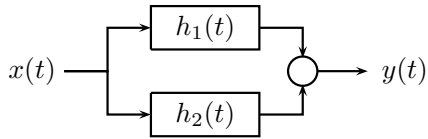


If $H_d(s)$ is minimum phase, we can simply choose $H_c(s) = \frac{1}{H_d(s)}$.
 If $H_d(s)$ is nonminimum phase, $H_c(s) = \frac{1}{H_d(s)}$ is unstable. To avoid instability, decompose: $H_d(s) = H_{d,min}(s)H_{d,ap}(s)$ and select:

$$H_c(s) = \frac{1}{H_{d,min}(s)} \implies \underbrace{H_d(s)H_c(s)}_{\text{magnitude distortion eliminated}} = H_{d,ap}(s)$$

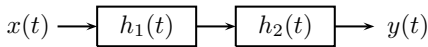
Transfer Functions of Interconnected LTI Systems

Section 9.8 in Oppenheim & Willsky



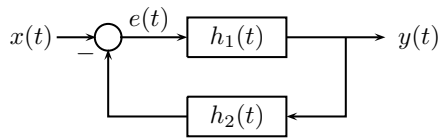
$$h(t) = h_1(t) + h_2(t)$$

$$H(s) = H_1(s) + H_2(s)$$



$$h(t) = h_1(t) * h_2(t)$$

$$H(s) = H_1(s)H_2(s)$$



$$E(s) = X(s) - H_2(s)Y(s)$$

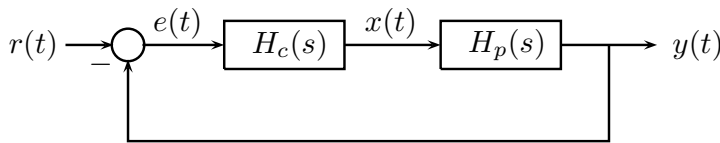
$$Y(s) = H_1(s)E(s)$$

$$= H_1(s)X(s) - H_1(s)H_2(s)Y(s)$$

$$(1 + H_1(s)H_2(s))Y(s) = H_1(s)X(s)$$

$$\frac{Y(s)}{X(s)} = H(s) = \frac{H_1(s)}{1 + H_1(s)H_2(s)}$$

Example: Feedback Control

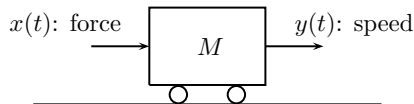


$r(t)$: reference signal to be tracked by $y(t)$

$H_c(s)$: controller, $H_p(s)$: system to be controlled - "plant"

$$H(s) = \frac{H_c(s)H_p(s)}{1 + H_c(s)H_p(s)}$$

Example:

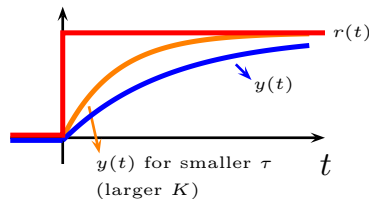


$$M \frac{dy}{dt} = x(t) \rightarrow MsY(s) = X(s)$$

$$H_p(s) = \frac{1}{Ms}$$

Take constant gain controller: $H_c(s) = K$

$$H(s) = \frac{\frac{K}{Ms}}{1 + \frac{K}{Ms}} = \frac{1}{\tau s + 1} \quad \tau = \frac{M}{K}$$



The Unilateral Laplace Transform

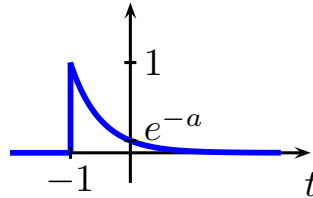
Section 9.9 in Oppenheim & Willsky

$$\mathcal{X}(s) = \int_{0^-}^{\infty} x(t)e^{-st} dt \quad (3)$$

Identical to the bilateral Laplace transform if $x(t) = 0$ for $t < 0$.

Example:

$$x(t) = e^{-a(t+1)}u(t+1)$$



$$X(s) = \frac{e^s}{s+a} \quad \text{Re}\{s\} > -a$$

$$\mathcal{X}(s) = \frac{e^{-a}}{s+a} \quad \text{Re}\{s\} > -a$$

Properties of the unilateral Laplace transform

Most properties of the bilateral Laplace transform also hold for the unilateral Laplace transform.

Exceptions:

Convolution:

$$x_1(t) * x_2(t) \longleftrightarrow \mathcal{X}_1(s)\mathcal{X}_2(s) \quad \text{if } x_1(t) = x_2(t) = 0 \text{ for all } t < 0$$

This follows from the convolution property of the bilateral Laplace transform which coincides with the unilateral transform because $x_1(t) = x_2(t) = 0$, $t < 0$.

Differentiation in Time:

$$\frac{dx(t)}{dt} \longleftrightarrow s\mathcal{X}(s) - x(0^-)$$

Repeated application gives:

$$\begin{aligned} \frac{d^2x(t)}{dt^2} &= \frac{d}{dt} \left\{ \frac{dx(t)}{dt} \right\} \longleftrightarrow s(s\mathcal{X}(s) - x(0^-)) - \frac{dx}{dt}(0^-) \\ &= s^2\mathcal{X}(s) - sx(0^-) - \frac{dx}{dt}(0^-) \\ \frac{d^3x(t)}{dt^3} &= \frac{d}{dt} \left\{ \frac{d^2x(t)}{dt^2} \right\} \longleftrightarrow s \left(s^2\mathcal{X}(s) - sx(0^-) - \frac{dx}{dt}(0^-) \right) - \frac{d^2x}{dt^2}(0^-) \\ &= s^3\mathcal{X}(s) - s^2x(0^-) - s\frac{dx}{dt}(0^-) - \frac{d^2x}{dt^2}(0^-) \end{aligned}$$

Solving differential equations with the unilateral Laplace transform

Example:

$$\frac{d^2 y(t)}{dt^2} + 3\frac{dy}{dt} + 2y(t) = e^t \quad t \geq 0 \quad (4)$$

Initial condition $y(0^-) = a$, $\frac{dy}{dt}(0^-) = b$.

$$(s^2 Y(s) - as - b) + 3(sY(s) - a) + 2Y(s) = \frac{1}{s-1}$$

$$(s^2 + 3s + 2)Y(s) = as + b + 3a + \frac{1}{s-1} = \frac{as^2 + (b+2a)s + (1-b-3a)}{s-1}$$

$$Y(s) = \frac{as^2 + (b+2a)s + (1-b-3a)}{(s+1)(s+2)(s-1)}$$

Partial fraction expansion:

$$\begin{aligned} Y(s) &= \frac{A_1}{s+1} + \frac{A_2}{s+2} + \frac{B}{s-1} \\ &= \frac{(A_1 + A_2 + B)s^2 + (A_1 + 3B)s + (2B - 2A_1 - A_2)}{(s+1)(s+2)(s-1)} \end{aligned}$$

Match coefficients:

$$\left. \begin{aligned} A_1 + A_2 + B &= a \\ A_1 + 3B &= b + 2a \\ 2B - 2A_1 - A_2 &= 1 - b - 3a \end{aligned} \right\} \begin{aligned} B &= 1/6 \\ A_1 &= -\frac{1}{2} + 2a + b \\ A_2 &= \frac{1}{3} - a - b \end{aligned}$$

Then,

$$y(t) = \frac{1}{6}e^t + \left(-\frac{1}{2} + 2a + b\right)e^{-t} + \left(\frac{1}{3} - a - b\right)e^{-2t} \quad t \geq 0.$$

Compare this to the standard method for solving linear constant coefficient differential equations:

The first term in $y(t)$ above is the particular solution. If we substitute $y_p(t) = \frac{1}{6}e^t$ in (4):

$$\frac{d^2 y_p(t)}{dt^2} + 3\frac{dy_p}{dt} + 2y_p(t) = e^t.$$

The second and third terms constitute the homogeneous solution. If we substitute $y_h(t) = A_1 e^{-t} + A_2 e^{-2t}$:

$$\frac{d^2 y_h(t)}{dt^2} + 3\frac{dy_h}{dt} + 2y_h(t) = 0.$$

Thus, $y(t) = y_p(t) + y_h(t)$ and A_1 and A_2 are selected to satisfy the initial conditions.