EE120 - Fall'15 - Lecture 1 Notes¹ Murat Arcak 26 August 2015

Linear Time-Invariant (LTI) Systems



Linearity: Two conditions must be satisfied:

1. Scaling:

$$ax(t) \rightarrow ay(t)$$
 for any number *a*; (1)

2. Superposition:

$$x_1(t) + x_2(t) \to y_1(t) + y_2(t).$$
 (2)

Corollary: If the input to a linear system is 0, the output must be 0. $\overrightarrow{Proof.}$ Choose a = 0 in the scaling property.

<u>Time-Invariance</u>: A time shift in the input results is an identical time shift in the output:

$$x(t-T) \to y(t-T). \tag{3}$$

Example: Moving average filter:

$$y[n] = \frac{1}{3} \left(x[n-1] + x[n] + x[n+1] \right) \to \text{LTI.}$$
 (4)

Example: Median Filter:

$$y[n] = med\{x[n-1], x[n], x[n+1]\} \rightarrow TI$$
, but nonlinear. (5)



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Discrete-Time (DT) LTI Systems: Convolution Sum

Section 2.1 in Oppenheim & Willsky

Let h[n] denote the response of an LTI system to the unit impulse:



Then, for any input x[n], the output is:

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] \quad \text{"convolution sum"}$$
(6)

Proof. Rewrite x[n] as

$$\begin{array}{rcl} x[n] & = & \ldots + x[-1]\delta[n+1] + x[0]\delta[n] + x[1]\delta[n-1] + \ldots & (7) \\ & & \\ & & \\ \end{array}$$

$$= \sum_{k=-\infty} x[k]\delta[n-k]$$
(8)

Since $\delta[n] \to h[n]$, by time-invariance: $\delta[n-k] \to h[n-k]$. Then, by linearity: $\sum_k x[k]\delta[n-k] \to \sum_k x[k]h[n-k]$.

Example: For the moving average system above,



$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

=
$$\sum_{k=n-1}^{n+1} \frac{1}{3}x[k] = \frac{1}{3}(x[n-1] + x[n] + x[n+1])$$
(9)

Example: For the median filter:



Since the system is nonlinear, we can't use convolution to predict the output.

Continuous-Time (CT) LTI Systems: Convolution Integral

Unit impulse:

$$\delta(t) \triangleq \lim_{\Delta \to 0} \delta_{\Delta}(t) \tag{10}$$

where $\delta_{\Delta}(t)$ is as in Figure 1.

Let h(t) denote the response of a LTI system to $\delta(t)$. Then, for any input x(t), the output is :

$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau)d\tau$$
 "convolution integral" (11)

Proof. First, note that the staircase approximation in Figure 2 recovers x(t) as $\Delta \rightarrow 0$:

$$x(t) = \lim_{\Delta \to 0} \sum_{k=-\infty}^{\infty} x(k\Delta) \Delta \delta_{\Delta}(t - k\Delta).$$
(12)

Next, let $h_{\Delta}(t)$ denote the response of the system to $\delta_{\Delta}(t)$ and note from the LTI property that the response to each term in the sum above is $x(k\Delta)\Delta h_{\Delta}(t-k\Delta)$. Thus, the response to x(t) is

$$y(t) = \lim_{\Delta \to 0} \sum_{k=-\infty}^{\infty} x(k\Delta) h_{\Delta}(t-k\Delta) \Delta = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau.$$
(13)

Section 2.2 in Oppenheim & Willsky





Figure 2: Staircase approximation of x(t).

Properties of LTI Systems

We will denote the convolution operation by "*". 1. Commutative Property:

$$x[n] * h[n] = h[n] * x[n]$$
 (14)

Proof.

$$\sum_{k=-\infty}^{\infty} x[k]h[n-k] = \sum_{r=-\infty}^{\infty} x[n-r]h[r],$$
(15)

with the change of variables $(n - k) \triangleq r$.

k

2. Distributive Property:

$$x[n] * (h_1[n] + h_2[n]) = x[n] * h_1[n] + x[n] * h_2[n]$$
(16)

$$x[n] \xrightarrow{h_1[n]} y[n] \equiv x[n] \xrightarrow{h_1[n] + h_2[n]} y[n]$$

3. Associative Property:

$$x[n] * (h_1[n] * h_2[n]) = (x[n] * h_1[n]) * h_2[n]$$
(17)

$$x[n] \longrightarrow h_1[n] \longrightarrow h_2[n] \longrightarrow y[n] \equiv x[n] \longrightarrow h_1[n] * h_2[n] \longrightarrow y[n]$$

Combine this with the commutative property

 $\longrightarrow h_1[n] \longrightarrow h_2[n] \longrightarrow = \longrightarrow h_2[n] \longrightarrow h_1[n] \longrightarrow$

Properties 1,2,3 above also hold for CT systems.

Section 2.3 in Oppenheim & Willsky

Determining Causality from the Impulse Response

For a DT LTI system, causality means:

$$h[n] = 0, \quad \forall n < 0. \tag{18}$$

For a CT LTI system, causality means:

$$h(t) = 0, \quad \forall t < 0. \tag{19}$$

Proof. Since $y[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k]$, if $h[k] \neq 0$ for some k < 0, then y[n] depends on x[n-k], where n-k > n.

Example: Moving average system above: $h[-1] \neq 0 \rightarrow$ noncausal.

Determining Stability from the Impulse Response

Stability criterion for a DT LTI system:

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty.$$
 (20)

Stability criterion for a CT LTI system:

$$\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty.$$
(21)

Proof.

Sufficiency: Suppose $\sum_{k=-\infty}^{\infty} |h[k]| < \infty$ and show that bounded inputs give bounded outputs:

 $|x[n]| \le B$ for all *n*, for some B > 0.

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$$|y[n]| = |\sum_k x[n-k]h[k]| \le \sum_k |x[n-k]| \cdot |h[k]| \le B \sum_k |h[k]| < \infty.$$

Necessity: To prove "stable $\Rightarrow \sum_k |h[k]| < \infty$ " prove the contrapositive:

$$\sum_{k} |h[k]| = \infty \Rightarrow \text{ unstable.}'' \tag{22}$$

Let $x[n] = sgn\{h[-n]\}$. Then, since $y[n] = \sum_{k=-\infty}^{\infty} h[k]x[n-k]$:

$$y[0] = \sum_{k=-\infty}^{\infty} h[k]x[-k] = \sum_{k} h[k]sign\{h[k]\} = \sum_{k} |h[k]| = \infty.$$
 (23)

Examples:

1. Moving average system above:

$$\sum_{k} |h[k]| = \frac{1}{3} + \frac{1}{3} + \frac{1}{3} = 1 \rightarrow \text{ stable.}$$
(24)

2. Integrator: $y(t) = \int_{-\infty}^{t} x(\tau) d\tau$. h(t) is the unit step (see Figure 3), and

$$\int_{-\infty}^{\infty} |h(\tau)| d\tau = \infty.$$
(25)

